



The Application of Vector Network Analyzers in Balanced Transmission Line Signal Integrity Measurements

Agenda

- **1. Introduction**
 - 1.1 The Application and Advantages of Balanced Transmission Lines
 - 1.2 Important Parameters and Characteristics for Balanced Transmission Lines
 - 1.3 Characteristic Impedance, Structural Return Loss (SRL), Attenuation, Delay, Skew, Near End Crosstalk (NEXT), Far End Crosstalk (FEXT)
 - 1.4 Single Ended, Common Mode, Differential Mode, Mixed Mode Scattering Parameters (S Parameters), Mode Conversion
- **2. Techniques Used to Connect Vector Network Analyzers (VNA) to Balanced Transmission Lines**
 - 2.1 Use of Baluns to Transform a 50 Ohm Un-balanced VNA Test Port to 100 Ohm Balanced Transmission Line
 - 2.2 Use of a Multiport VNA Test Set to Transform 50 Ohm Un-balanced VNA Test Ports to 100 Ohm Balanced Transmission Lines
 - 2.3 Advantages and Limitations of Each Approach
- **3. Frequency Domain Measurements for Balanced Transmission Lines**
 - 3.1 Frequency Domain Measurement with a VNA
 - 3.2 Frequency Domain Measurement with a TDR / TDT (FFT)
 - 3.3 Advantages and Limitations of Each Approach

Agenda

- **4. Time Domain Characteristics and Parameters for Balanced Transmission Lines**
 - 4.1 Characteristic Impedance, Fault Type and Location, Attenuation, Delay, Skew, Near End Crosstalk (NEXT), Far End Crosstalk (FEXT)
 - 4.2 Single Ended, Differential Mode, Common Mode, Mixed Mode, Mode Conversion
 - 4.3 Time Domain Measurement with a VNA (IFT)
 - 4.4 Time Domain Measurement with a TDR / TDT
 - 4.5 Advantages and Limitations of Each Approach
- **5. Techniques Used with MultiPort Vector Network Analyzers (VNA) for Balanced Transmission Line Measurements**
 - 5.1 Use of Mathematical Superposition with a Single Source MultiPort VNA for Balanced Transmission Line Measurements
 - 5.2 Use of Dual Differential Sources MultiPort VNA for Balanced Transmission Line Measurements
 - 5.3 Advantages and Limitations of Each Approach
- **6. Fixture and Launch Considerations in Connecting to Balanced Structures**

Introduction

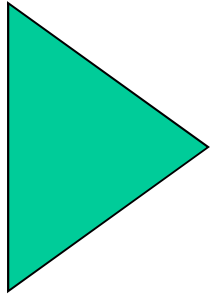
What is Signal Integrity (SI)?

- An Engineering Practice
 - That ensures all signals transmitted are received correctly
 - That ensures signals don't interfere with one another in a way to degrade reception.
 - That ensures signals don't damage any devices
 - That ensures signals don't pollute the electromagnetic spectrum

Introduction

Components of High Speed Design

✓ Transmitter



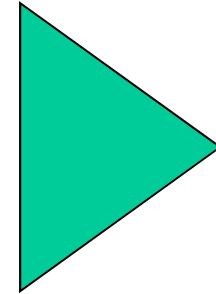
- Transistors
- Sources
- Algorithms
- Passives
- Memory

✓ Interconnect



- Circuit elements
- Transmission lines
- S - parameter blocks

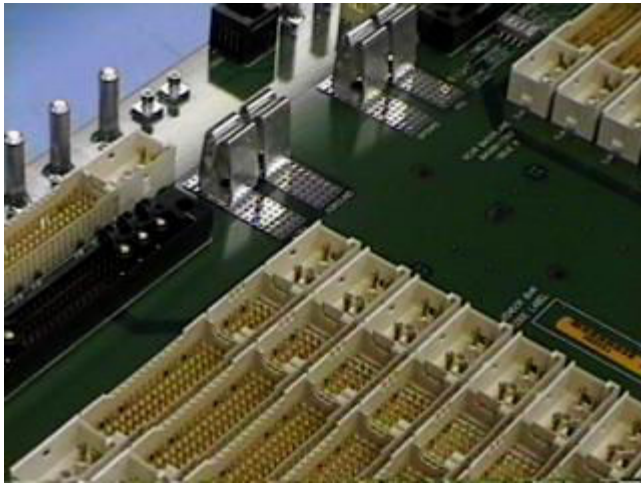
✓ Receiver



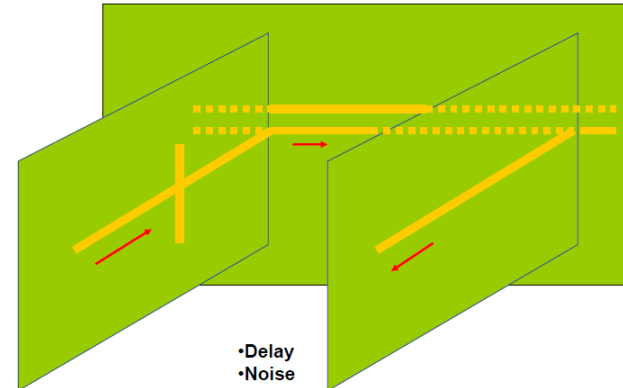
- Transistors
- Passives
- Algorithms
- Memory

Introduction

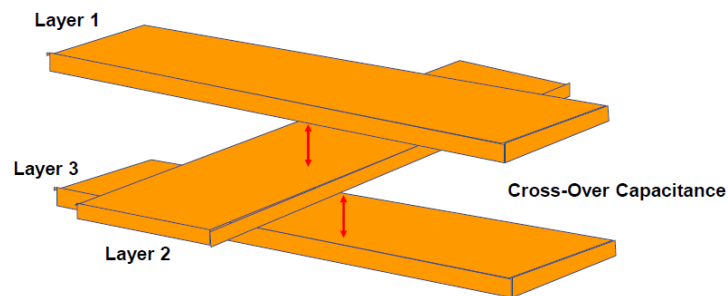
Components of High Speed Design



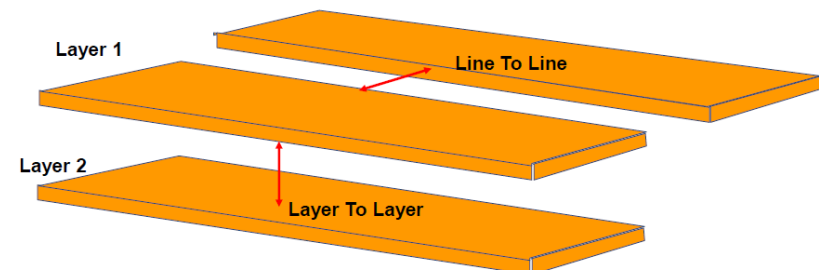
SIGNAL INTEGRITY



SIGNAL INTEGRITY
IMPEDANCE CHANGE
Cross-Over Capacitance



SIGNAL INTEGRITY
CROSSTALK

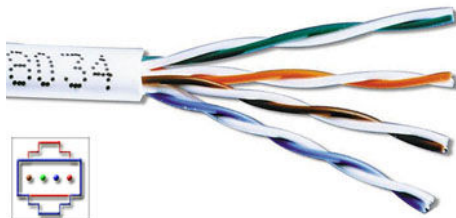


The Application and Advantages of Balanced Transmission Lines

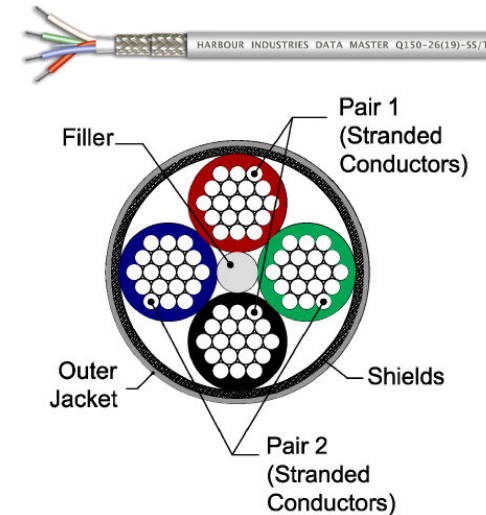
Shielded twisted pair (STP)



Unshielded twisted pair (UTP)

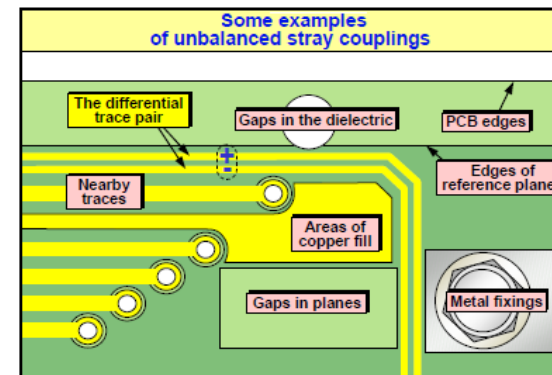
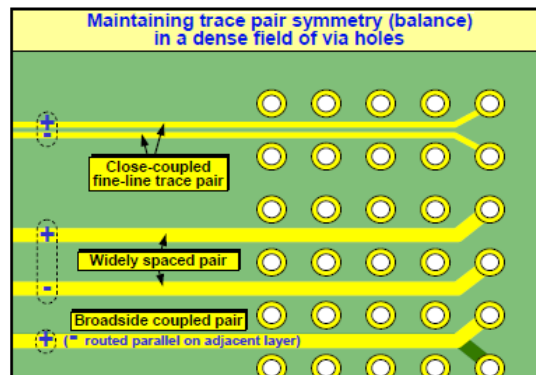
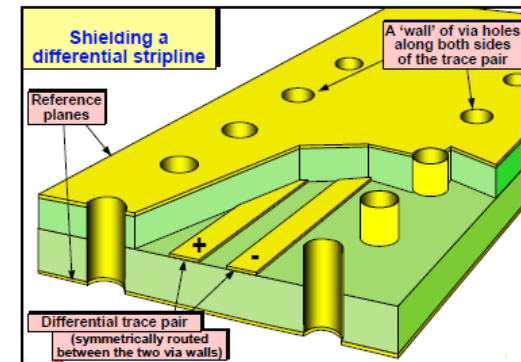
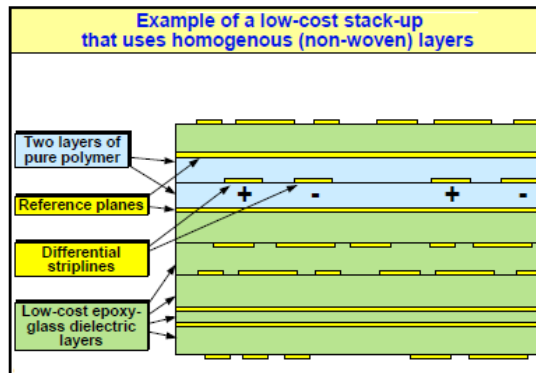


Shielded Quad Data Cable (Fiber-Channel)



Some examples of the types of differential transmission lines that can be constructed on PCBs					
Differential microstrip (edge-coupled)	Differential microstrip (edge-coupled) with plane	Coated differential microstrip (edge-coupled)	Coated diff. microstrip (edge-coupled) with plane	Embedded differential microstrip (edge-coupled)	Embedded diff. microstrip (edge-coupled) with plane
Coplanar differential microstrip	Coplanar differential microstrip with plane	Coated coplanar differential microstrip	Coated coplanar diff. microstrip with plane	Embedded coplanar differential microstrip	Embedded coplanar diff. microstrip with plane
Symmetrical differential stripline (edge-coupled)	Offset differential stripline (edge-coupled)	Broadside-coupled differential stripline	Broadside-coupled offset diff. stripline	Symmetrical coplanar differential stripline	Offset coplanar differential stripline

The Application and Advantages of Balanced Transmission Lines



Important Parameters and Characteristics for Balanced Transmission Lines

Two, single ended transmission lines with coupling



One, differential pair



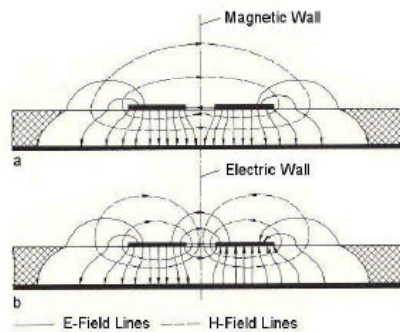
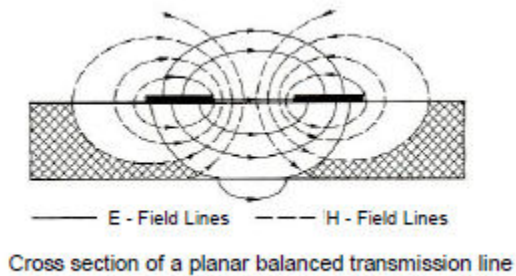
Two equivalent views.



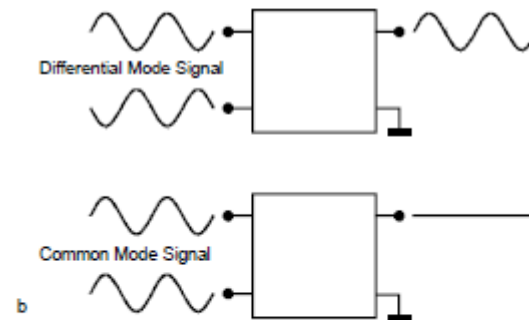
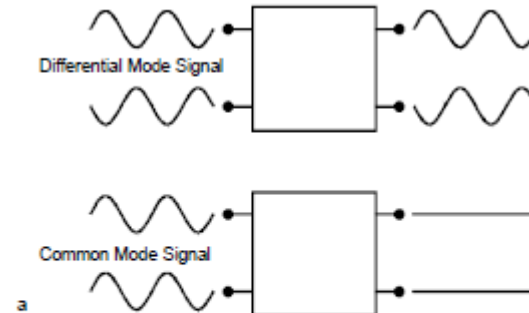
			Stimulus			
			Differential signal		Common signal	
			port 1	port 2	port 1	port 2
Response	Differential signal	port 1	SDD11	SDD12	SDC11	SDC12
		port 2	SDD21	SDD22	SDC21	SDC22
	Common signal	port 1	SCD11	SCD12	SCC11	SCC12
		port 2	SCD21	SCD22	SCC21	SCC22

Differential S parameters

Important Parameters and Characteristics for Balanced Transmission Lines



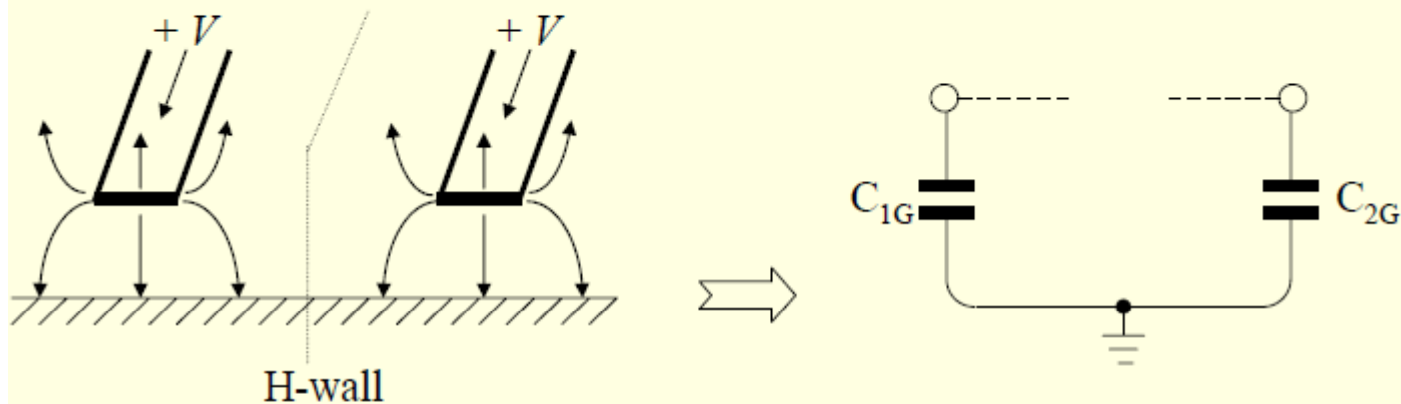
Cross section of a planar balanced transmission line, with electric and magnetic field lines
a: Even or common mode
b: Odd or differential mode



Ideal balanced devices:
a: Fully balanced
b: Balanced-to-single-ended

Important Parameters and Characteristics for Balanced Transmission Lines

Even Mode (Field)

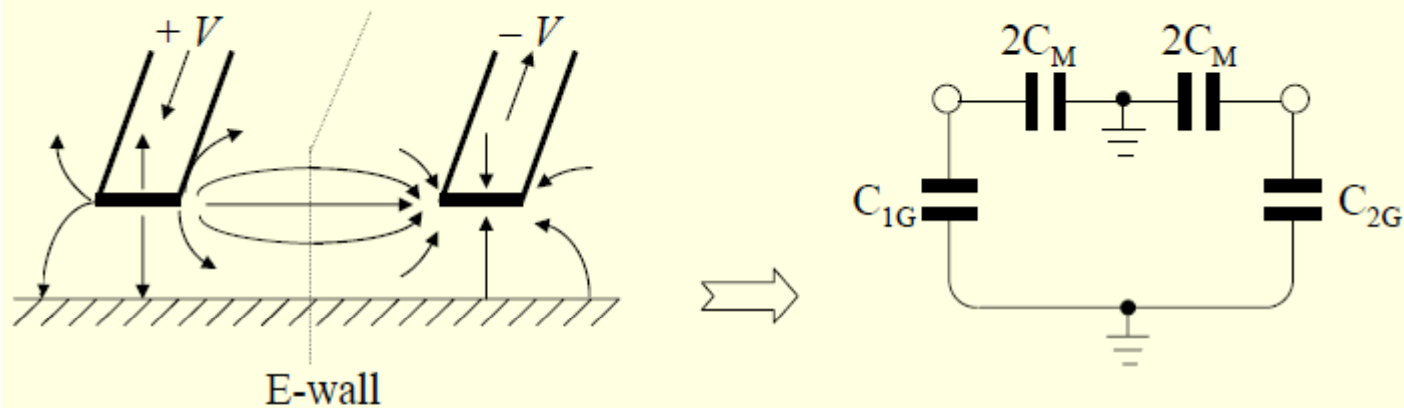


$$Z_{0e} = \sqrt{\frac{L_e}{C_e}} = \sqrt{\frac{L + L_M}{C_{1G}}} = \sqrt{\frac{L + L_M}{C_{2G}}}$$

$$v_{p,e} = \sqrt{\frac{1}{L_e C_e}} \left\{ = \frac{c}{\sqrt{\epsilon_r}} \right\} \text{ for homogeneous medium}$$

Important Parameters and Characteristics for Balanced Transmission Lines

Odd Mode (Field)

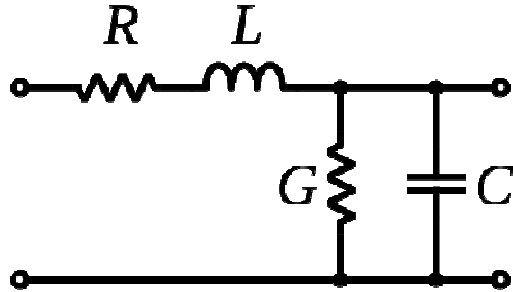


$$Z_{0o} = \sqrt{\frac{L_o}{C_o}} = \sqrt{\frac{L - L_M}{C_{1G} + 2C_M}} = \sqrt{\frac{L - L_M}{C_{2G} + 2C_M}}$$

$$v_{p,o} = \sqrt{\frac{1}{L_o C_o}} \left\{ = \frac{c}{\sqrt{\epsilon_r}} \right\} \text{ for homogeneous medium}$$

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Important Parameters and Characteristics for Balanced Transmission Lines



Propagation equation

$$\gamma = \sqrt{(R + j\omega L)(G + j\omega C)} = \alpha + j\beta$$

α is the attenuation (loss) factor

β is the phase (velocity) factor

Characteristic Impedance equation

$$Z_0 = \sqrt{\frac{R + j\omega L}{G + j\omega C}}$$

- **Propagation delay per unit length (T_0)**
{time/distance} [ps/in]
 - **Or Velocity (v_0) {distance/time} [in/ps]**
- **Characteristic Impedance (Z_0)**
- **Per-unit-length Capacitance (C_0) [pF/in]**
- **Per-unit-length Inductance (L_0) [nH/in]**
- **Per-unit-length (Series) Resistance (R_0)**
[W/in]
- **Per-unit-length (Parallel) Conductance (G_0)**
[S/in]

Important Parameters and Characteristics for Balanced Transmission Lines

- Knowing any two out of Z_0 , T_d , C_0 , and L_0 , the other two can be calculated.
- C_0 and L_0 are reciprocal functions of the line cross-sectional dimensions and are related by constant $\mu\epsilon$.
- ϵ is electric permittivity
 - $\epsilon_0 = 8.85 \times 10^{-12} \text{ F/m}$ (free space)
 - ϵ_r is relative dielectric constant
- μ is magnetic permeability
 - $\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$ (free space)
 - μ_r is relative permeability

$$Z_0 = \sqrt{\frac{L_0}{C_0}}$$

$$T_d = \sqrt{L_0 C_0}$$

$$C_0 = \frac{T_0}{Z_0}$$

$$L_0 = Z_0 T_0$$

$$v_0 = \frac{1}{\sqrt{\mu\epsilon}}$$

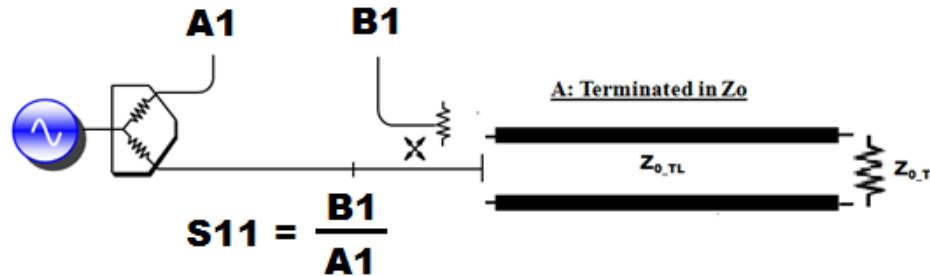
$$C_0 L_0 = \mu\epsilon$$

$$\mu = \mu_r \mu_0$$

$$\epsilon = \epsilon_r \epsilon_0$$

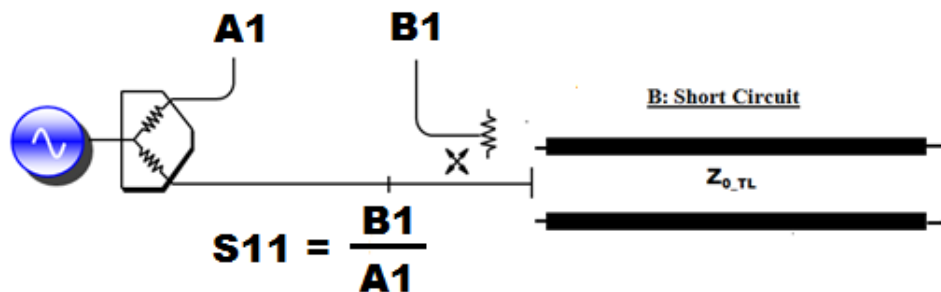
Important Parameters and Characteristics for Balanced Transmission Lines

Measurement of Characteristic Impedance Z_0



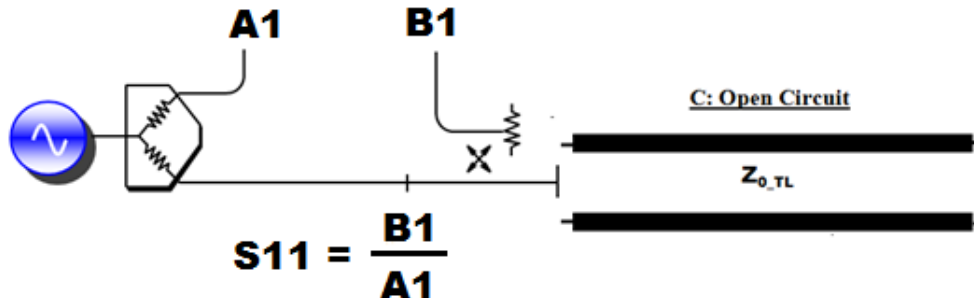
Measure S_{11} with Far End Terminated in Z_0

$$Z_{0_TL} = \frac{1 + S_{11}}{1 - S_{11}} * Z_0$$



Measure S_{11} with Far End Terminated in Short Circuit

$$Z_{oc} = \frac{1 + S_{11}}{1 - S_{11}} * Z_0$$



Measure S_{11} with Far End Terminated in Open Circuit

$$Z_{sc} = \frac{1 + S_{11}}{1 - S_{11}} * Z_0$$

$$Z_{0_TL} = \sqrt{Z_{oc} * Z_{sc}} \quad 15$$

Important Parameters and Characteristics for Balanced Transmission Lines

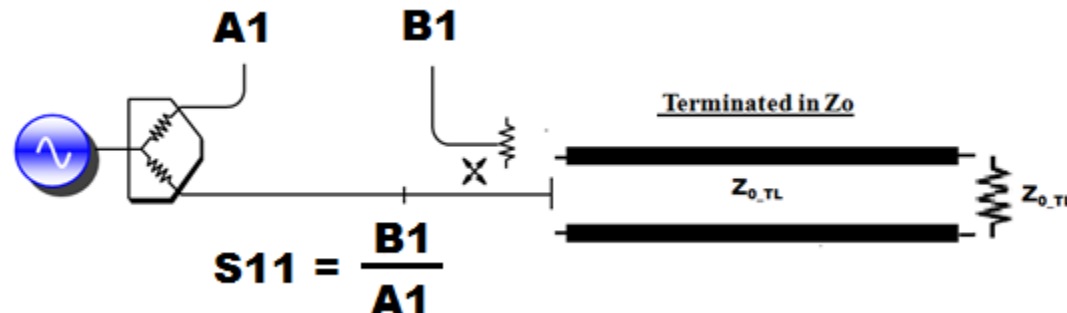
Measurement of Reflection Characteristics

Reflection Coefficient = $\rho = S_{D1D1} = b_1 / a_1 = \text{reflected} / \text{incident}$

Voltage Standing Wave Ratio = VSWR =
$$\frac{1 + |S_{D1D1}|}{1 - |S_{D1D1}|}$$

Return Loss (dB) = $-20 \log_{10} (|S_{D1D1}|)$

[Structural Return Loss (dB)]



Important Parameters and Characteristics for Balanced Transmission Lines

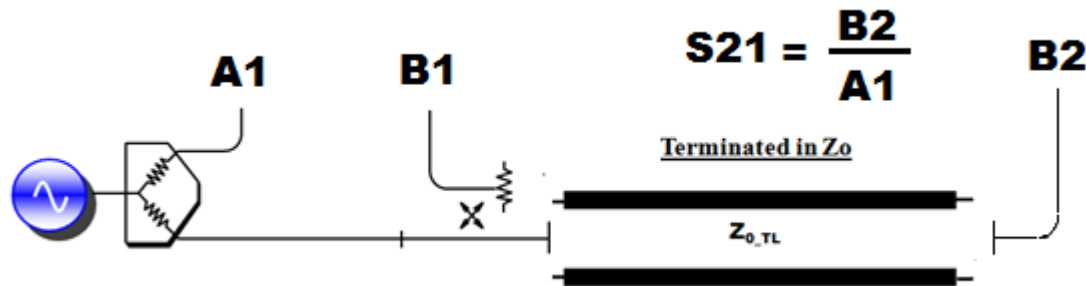
Measurement of Transmission Characteristics

Single Pair Measurements

$$\text{Insertion Loss (dB)} = 20 \log_{10} (| S_{D2D1} |) [\text{Log Magnitude}]$$

$$\text{Attenuation (dB/in)} = 20 \log_{10} (| S_{D2D1} |) / (\text{Length of Transmission Line})$$

$$\text{Delay (ns)} = (S_{D2D1}) [\text{Group Delay}]$$

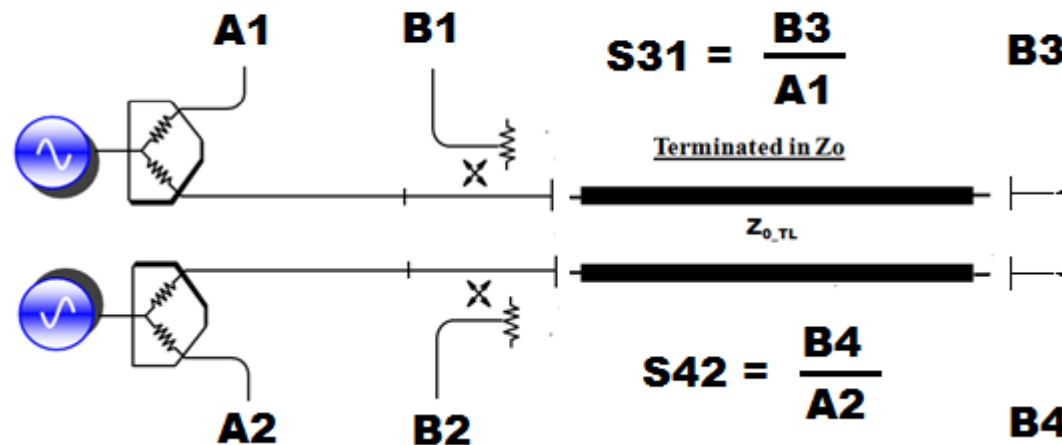


Important Parameters and Characteristics for Balanced Transmission Lines

Measurement of Transmission Characteristics

Single Pair Measurements

$$\text{Delay Skew (ns)} = (S31) [\text{Group Delay}] - (S42) [\text{Group Delay}]$$

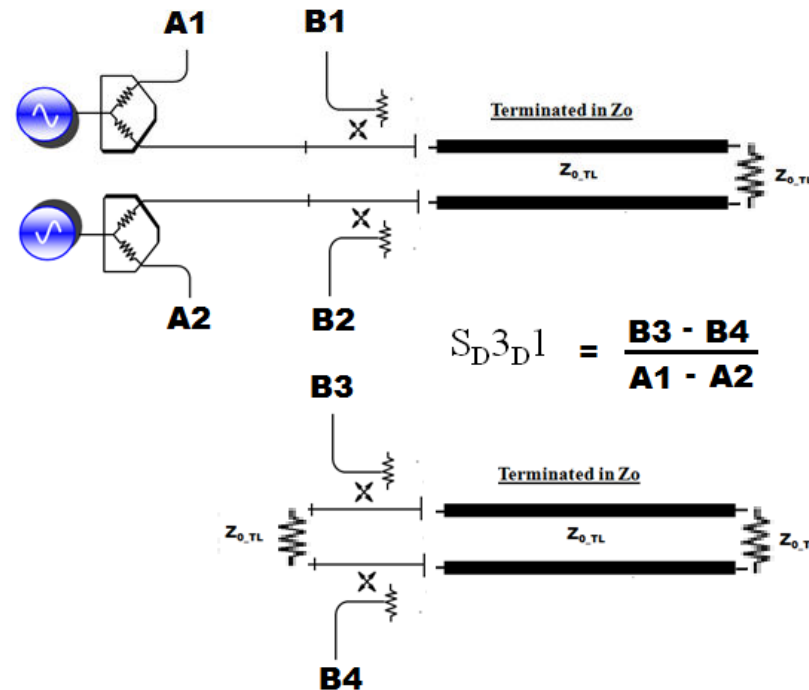


Important Parameters and Characteristics for Balanced Transmission Lines

Measurement of Transmission Characteristics

Two Pair Measurements

Near End Crosstalk (NEXT) (dB) = $(S_D 3_{D1})$ [Log Magnitude]

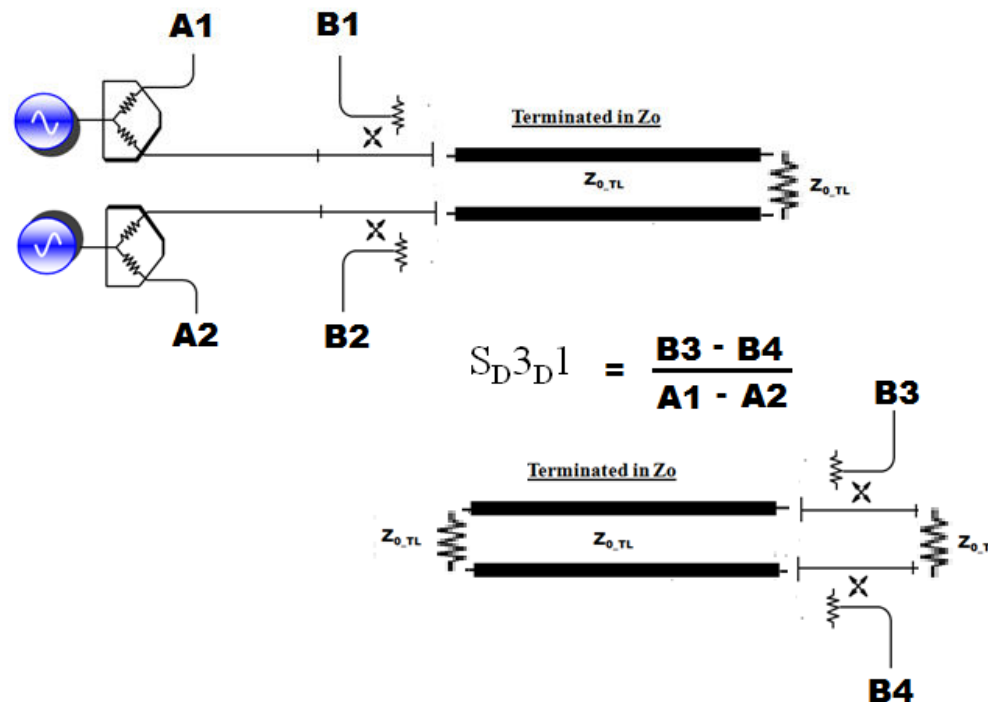


Important Parameters and Characteristics for Balanced Transmission Lines

Measurement of Transmission Characteristics

Two Pair Measurements

Far End Crosstalk (FEXT) (dB) = (S_{D3D1}) [Log Magnitude]



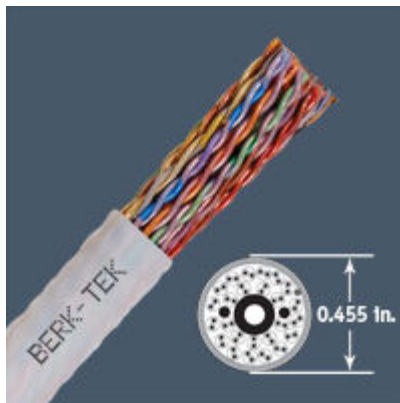
Important Parameters and Characteristics for Balanced Transmission Lines

Measurement of Transmission Characteristics

Multiple Pair Measurements

Power Sum - NEXT (dB) = Σ NEXT Reference Pair to All Other Pairs

Power Sum - ELFEXT (dB) = Σ FEXT Reference Pair to All Other Pairs



TECHNICAL DATA ELECTRICAL

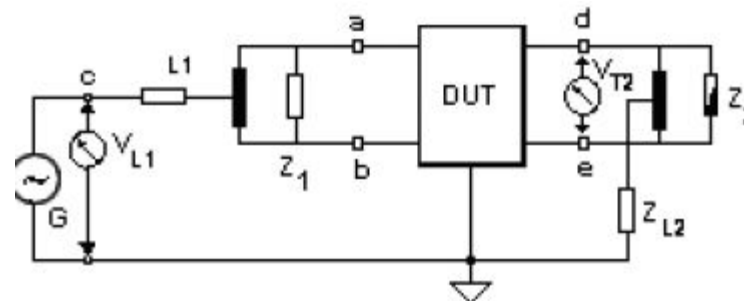
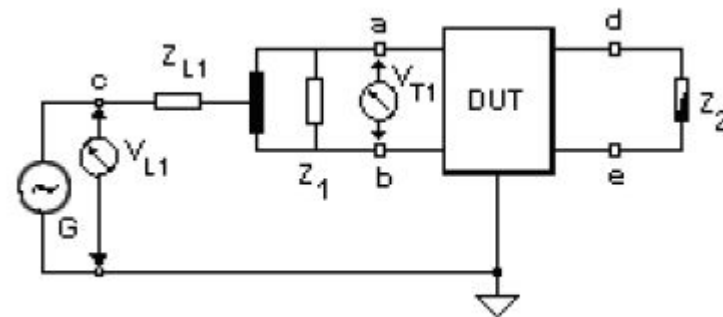
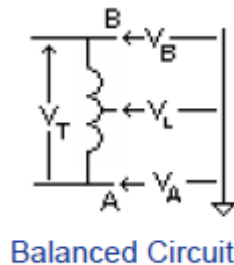
FREQUENCY (MHz)	Insertion Loss (dB/100 m max.)	RL (dB min.)	SRL (dB min.)	PS-NEXT (dB min.)	NEXT (dB min.)	ELFEXT (dB min.)	PS-ELFEXT (dB min.)
1	2	20	25	62.3	65.3	63.8	60.8
4	4.1	23	25	53.2	56.2	51.7	48.7
10	6.5	25	25	47.3	50.3	43.8	40.8
16	8.2	25	25	44.2	47.3	39.7	36.7
20	9.3	25	25	42.7	45.8	37.7	34.7
31.25	11.7	23.6	23.6	39.8	42.9	33.9	30.9
62.5	17	21.5	21.5	35.3	38.4	27.8	24.8
100	22	20.1	20.1	32.3	35.3	23.8	20.8
150	27.5	18.9	18.9	29.7	32.7	20.2	17.2
200	32.4	18.0	18.0	27.8	30.8	17.7	14.7
250	36.9	17.3	17.3	26.3	29.3	15.8	12.8

IMPORTANT: Berk-Tek performance guarantees are based on swept-frequency testing and apply to all frequencies for the entire specified frequency range and are not limited to the tables of data shown which are presented to demonstrate our guarantees at "representative" frequencies. Values above 100 MHz are for engineering information. Other jacket colors available.

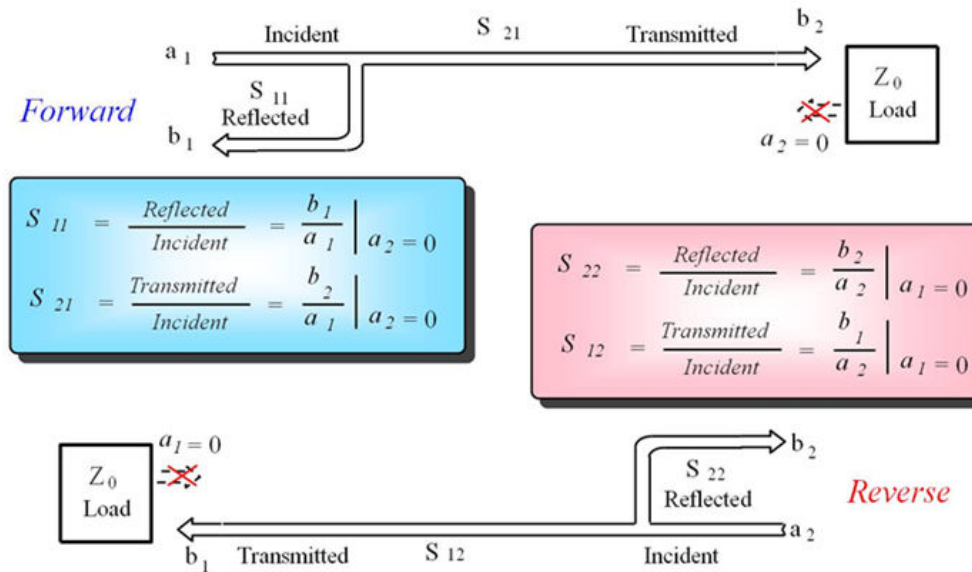
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Important Parameters and Characteristics for Balanced Transmission Lines

Measurement of Longitudinal Balance



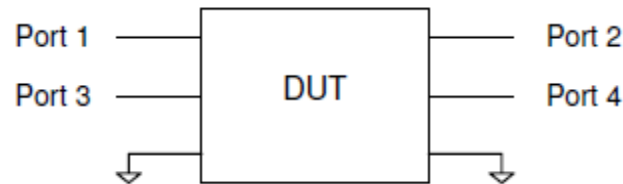
Important Parameters and Characteristics for Balanced Transmission Lines Single Ended S-Parameters



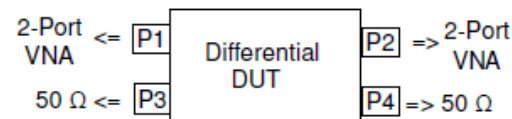
- Traditional S-Parameters
- All Measurements Use Ground as the Return (Common Mode)
- Balanced Lines and Devices Measured One Side at a Time
- Delay Skew is Based on Single Ended Measurements

Important Parameters and Characteristics for Balanced Transmission Lines

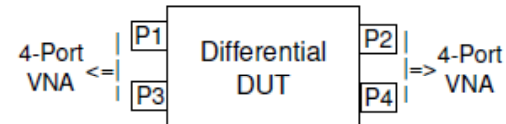
Common Mode S-Parameters



$$\begin{bmatrix} b_1 \\ b_2 \\ b_3 \\ b_4 \end{bmatrix} = \begin{bmatrix} S_{11} & S_{12} & S_{13} & S_{14} \\ S_{21} & S_{22} & S_{23} & S_{24} \\ S_{31} & S_{32} & S_{33} & S_{34} \\ S_{41} & S_{42} & S_{43} & S_{44} \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \\ a_3 \\ a_4 \end{bmatrix}$$



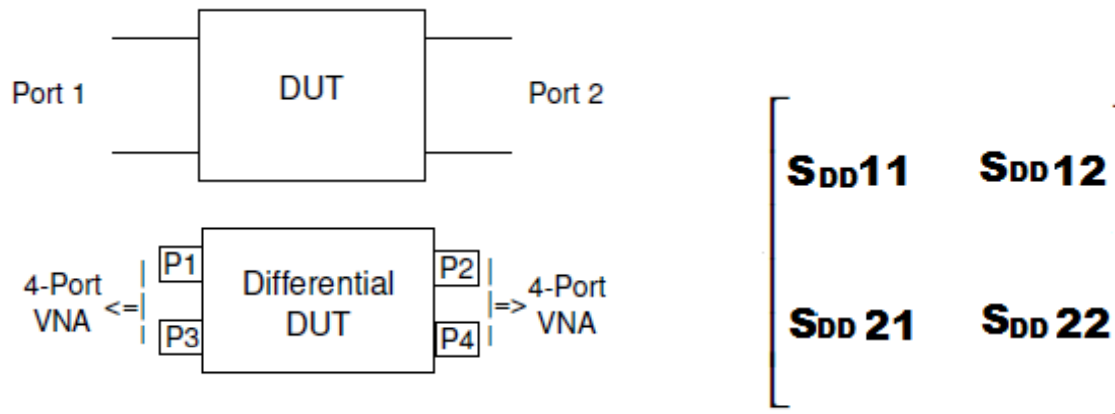
For single-ended 2-port VNA.



For differential-ended 4-port VNA.

- Balanced Transmission Lines / Devices Treated as a 4 Port
- Ground is Common Return For All Ports (May be Virtual)
- Each DUT “Port” is Connected to a VNA Port Center Conductor
- Delay Skew Can Be Derived From Common Mode Measurements

Important Parameters and Characteristics for Balanced Transmission Lines Differential Mode S-Parameters



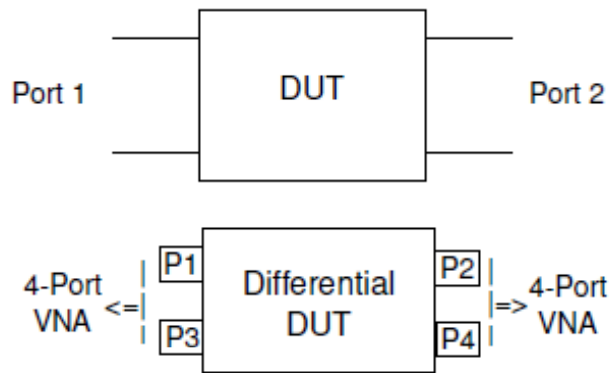
For differential-ended 4-port VNA.

- Balanced Transmission Lines / Devices Treated as a 4 Port
- Ground is Common Return For All Ports (May be Virtual)
- Each DUT “Port” is Connected to a VNA Port Center Conductor
- Delay Skew Can Be Derived From Common Mode Measurements

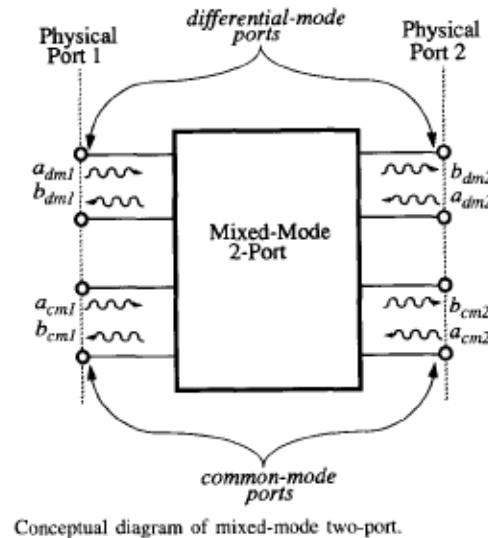
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Important Parameters and Characteristics for Balanced Transmission Lines

Mixed Mode S-Parameters



For differential-ended 4-port VNA.



Conceptual diagram of mixed-mode two-port.

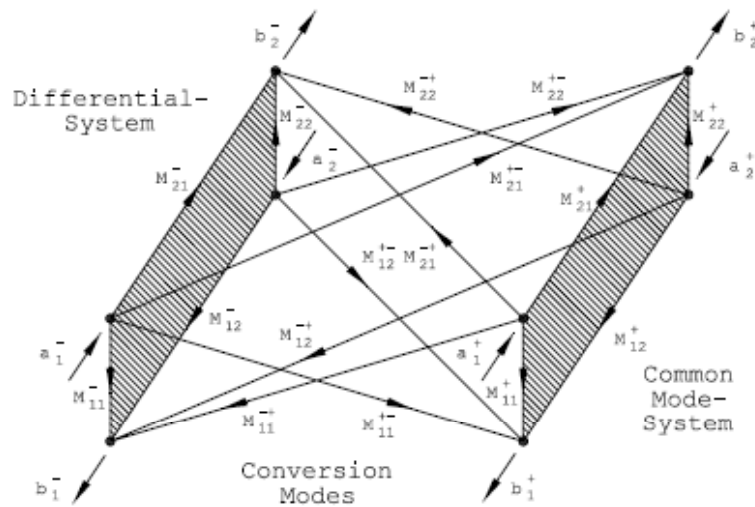
$$\begin{bmatrix} a_{d1} \\ a_{d2} \\ a_{c1} \\ a_{c2} \end{bmatrix} = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & -1 & 0 & 0 \\ 0 & 0 & 1 & -1 \\ 1 & 1 & 0 & 0 \\ 0 & 0 & 1 & 1 \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \\ a_3 \\ a_4 \end{bmatrix}$$

$$\begin{bmatrix} b_{d1} \\ b_{d2} \\ b_{c1} \\ b_{c2} \end{bmatrix} = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & -1 & 0 & 0 \\ 0 & 0 & 1 & -1 \\ 1 & 1 & 0 & 0 \\ 0 & 0 & 1 & 1 \end{bmatrix} \begin{bmatrix} b_1 \\ b_2 \\ b_3 \\ b_4 \end{bmatrix}$$

$$\begin{bmatrix} b_{d1} \\ b_{d2} \\ b_{c1} \\ b_{c2} \end{bmatrix} = \begin{bmatrix} S_{dd1} & S_{dd2} \\ S_{dd21} & S_{dd22} \\ S_{cd1} & S_{cd2} \\ S_{cd21} & S_{cd22} \end{bmatrix} \begin{bmatrix} S_{dd1} & S_{dd2} \\ S_{dd21} & S_{dd22} \\ S_{cd1} & S_{cd2} \\ S_{cd21} & S_{cd22} \end{bmatrix} \begin{bmatrix} a_{d1} \\ a_{d2} \\ a_{c1} \\ a_{c2} \end{bmatrix}$$

- Balanced Transmission Lines / Devices Treated as a 4 Port
- Ground is Common Return For All Ports (May be Virtual)
- Each DUT “Port” is Connected to a VNA Port Center Conductor
- Mixed Mode Parameters Include Important Mode Conversion Parameters , Differential Mode ↔ Common Mode

Important Parameters and Characteristics for Balanced Transmission Lines Mode Conversion



Signal flow diagram of the multi-mode parameters

M-parameters for the differential mode:

$$\begin{aligned} M_{11}^{-} &= \frac{1}{2} (S_{11} + S_{22} - S_{12} - S_{21}) \\ M_{12}^{-} &= \frac{1}{2} (S_{13} + S_{24} - S_{14} - S_{23}) \\ M_{21}^{-} &= \frac{1}{2} (S_{31} + S_{42} - S_{41} - S_{32}) \\ M_{22}^{-} &= \frac{1}{2} (S_{33} + S_{44} - S_{34} - S_{43}) \end{aligned}$$

M-parameters for the common mode:

$$\begin{aligned} M_{11}^{+} &= \frac{1}{2} (S_{11} + S_{22} + S_{12} + S_{21}) \\ M_{12}^{+} &= \frac{1}{2} (S_{13} + S_{24} + S_{14} + S_{23}) \\ M_{21}^{+} &= \frac{1}{2} (S_{31} + S_{42} + S_{41} + S_{32}) \\ M_{22}^{+} &= \frac{1}{2} (S_{33} + S_{44} + S_{34} + S_{43}) \end{aligned}$$

Conversion parameters for the differential into the common mode:

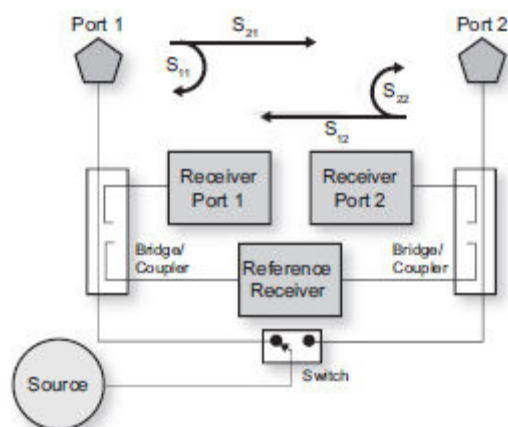
$$\begin{aligned} M_{11}^{+-} &= \frac{1}{2} (S_{11} - S_{22} - S_{12} + S_{21}) \\ M_{12}^{+-} &= \frac{1}{2} (S_{13} - S_{24} - S_{14} + S_{23}) \\ M_{21}^{+-} &= \frac{1}{2} (S_{31} - S_{42} + S_{41} - S_{32}) \\ M_{22}^{+-} &= \frac{1}{2} (S_{33} - S_{44} - S_{34} + S_{43}) \end{aligned}$$

Conversion parameters for common into the differential mode:

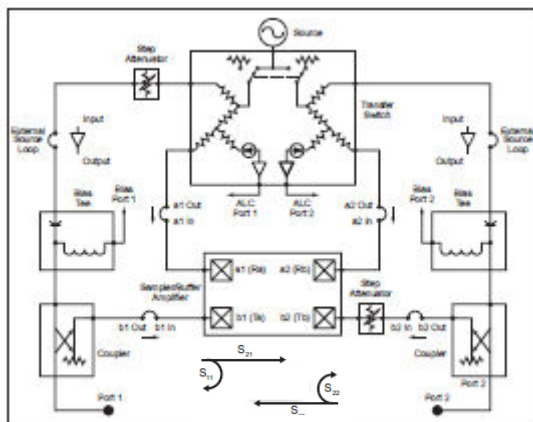
$$\begin{aligned} M_{11}^{-+} &= \frac{1}{2} (S_{11} - S_{22} + S_{12} - S_{21}) \\ M_{12}^{-+} &= \frac{1}{2} (S_{13} - S_{24} + S_{14} - S_{23}) \\ M_{21}^{-+} &= \frac{1}{2} (S_{31} - S_{42} - S_{41} + S_{32}) \\ M_{22}^{-+} &= \frac{1}{2} (S_{33} - S_{44} + S_{34} - S_{43}) \end{aligned}$$

- Balanced Transmission Lines / Devices Treated as a 4 Port
- Ground is Common Return For All Ports (May be Virtual)
- Each DUT “Port” is Connected to a VNA Port Center Conductor
- Mixed Mode Parameters Include Important Mode Conversion Parameters , Differential Mode ↔ Common Mode

Techniques Used to Connect Vector Network Analyzers (VNA) to Balanced Transmission Lines



Typical 2 Port VNA (3 Receiver)



Typical 2 Port VNA (4 Receiver)

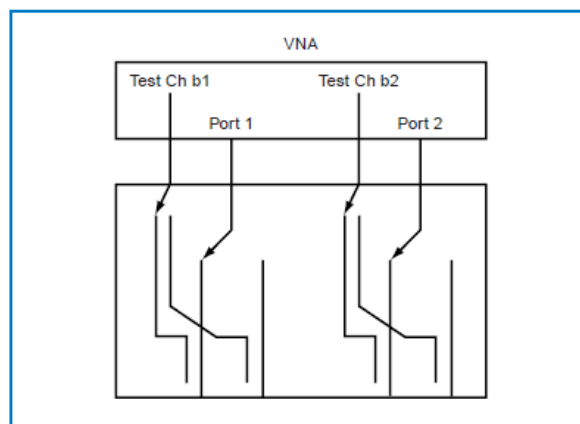


Figure 1. A coupler test set architecture is shown here; the test couplers are on the DUT-side of the multiplexing switches. As N becomes large, this test set becomes quite complex.

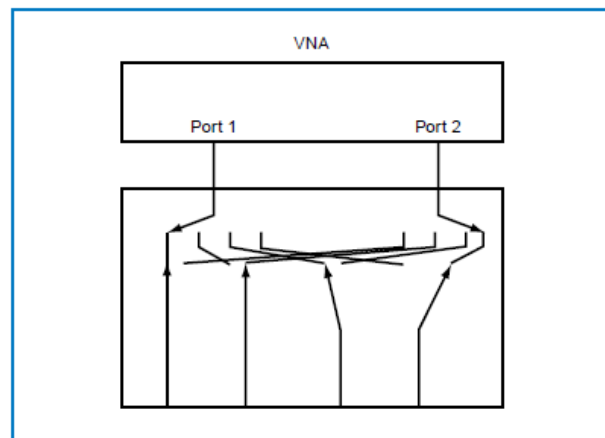
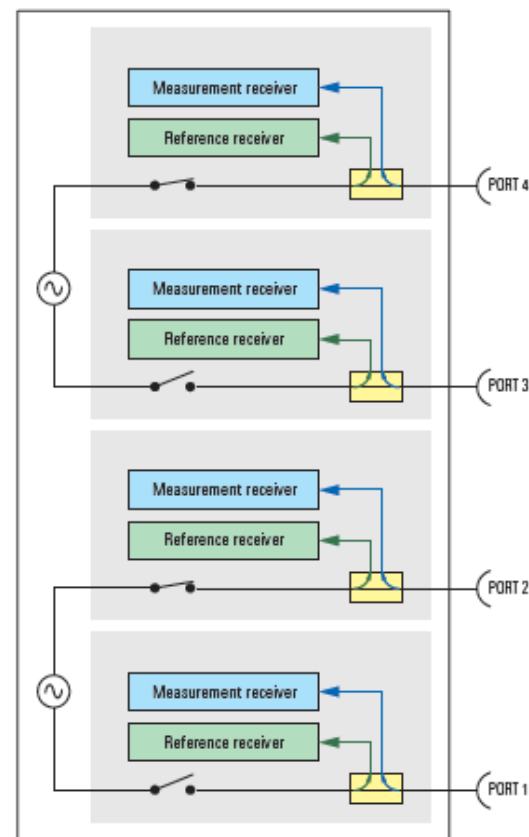
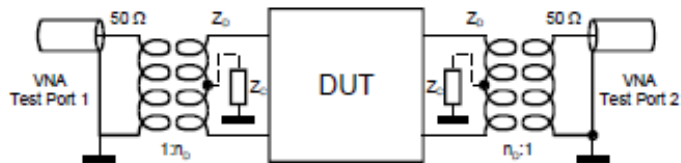


Figure 2. A no-coupler architecture for the 4x2 problem is shown here. In this case, any VNA port can be connected to any test port although this is not needed for most measurements. This test set can be simpler for large N but does have limitations.

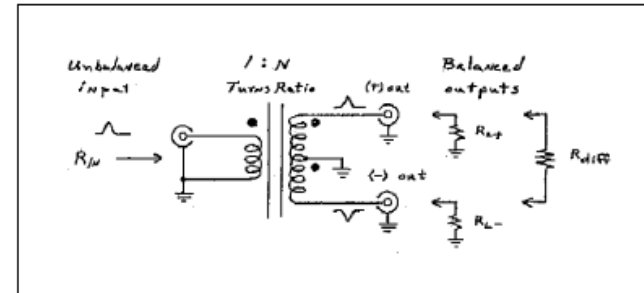


Typical 4 Port VNA (2 Source)

Use of Baluns to Transform a 50 Ohm Un-balanced VNA Test Port to 100 Ohm Balanced Transmission Line



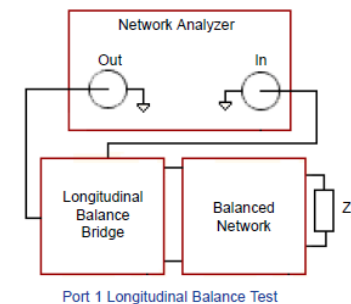
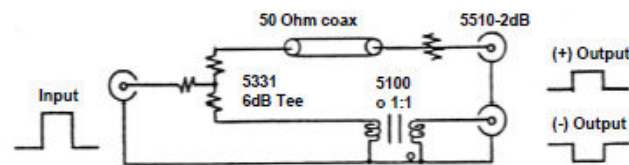
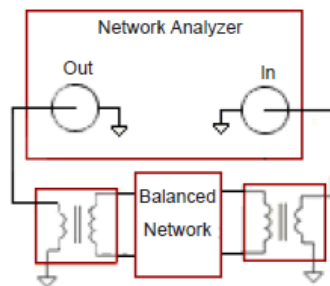
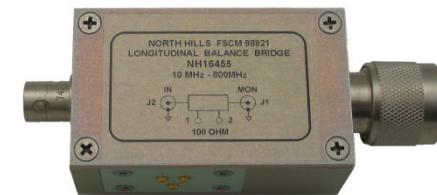
Measuring the differential mode characteristics of a two-port DUT using baluns



BALUN Transformer with Center-Tapped Output



Picosecond Pulse Labs Model 5315A BALUN
17 GHz typical BW



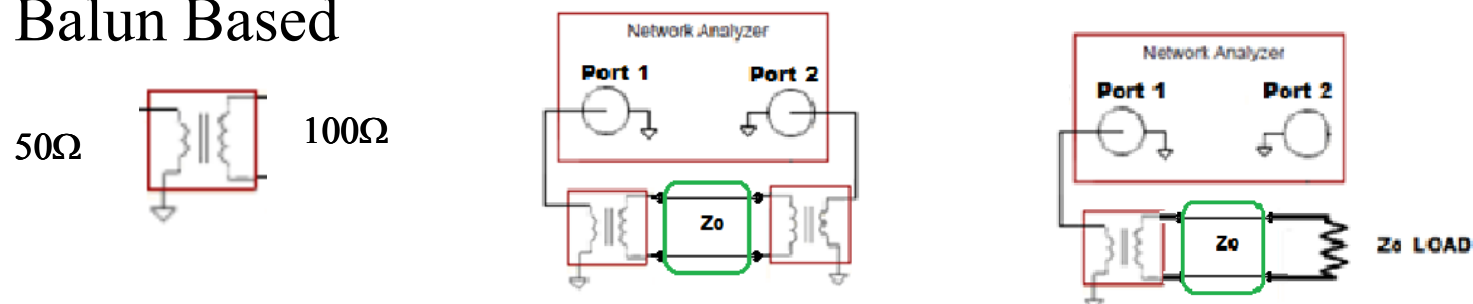
Port 1 Longitudinal Balance Test

Use of Baluns to Transform a 50 Ohm Un-balanced VNA Test Port to 100 Ohm Balanced Transmission Line

VNA Configurations for Balun Based Measurements

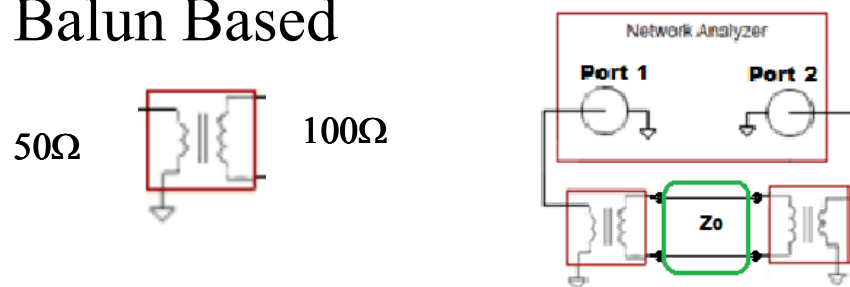
Reflection Measurements

Balun Based



Transmission Measurements

Balun Based

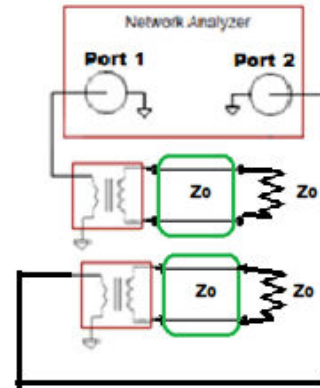
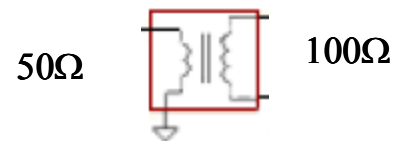


Use of Baluns to Transform a 50 Ohm Un-balanced VNA Test Port to 100 Ohm Balanced Transmission Line

VNA Configurations for Balun Based Measurements

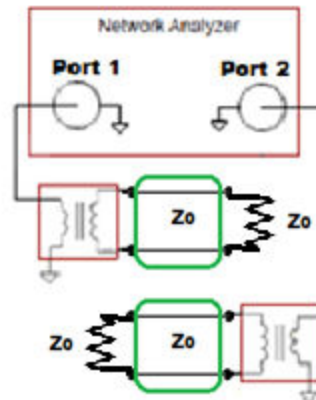
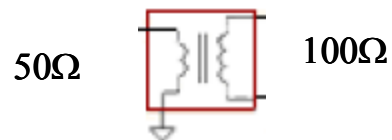
NEXT Measurements

Balun Based



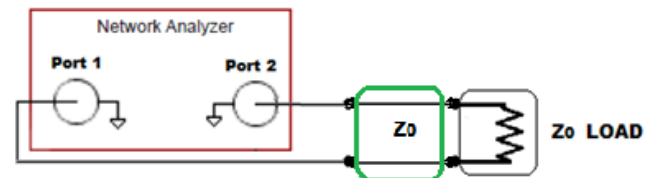
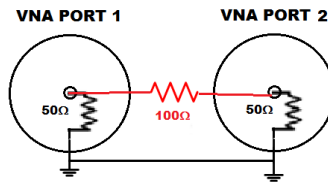
FEXT Measurements

Balun Based



Use of a 2 Port VNA Test Set to Transform 50 Ohm Un-balanced VNA Test Ports to 100 Ohm Balanced Transmission Lines VNA Configurations for Reflection Measurements

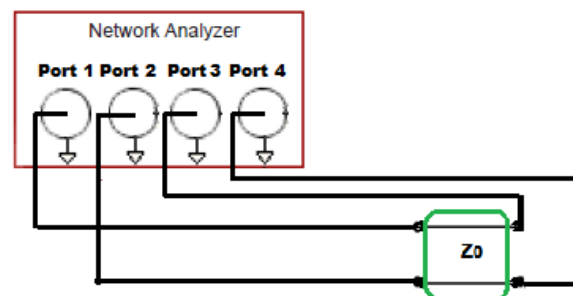
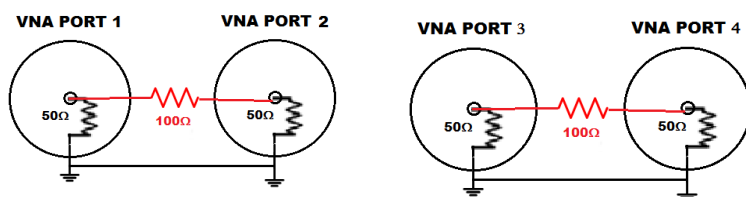
Multiport (2 Port) Based



Use of a Multiport VNA Test Set to Transform 50 Ohm Un-balanced VNA Test Ports to 100 Ohm Balanced Transmission Lines

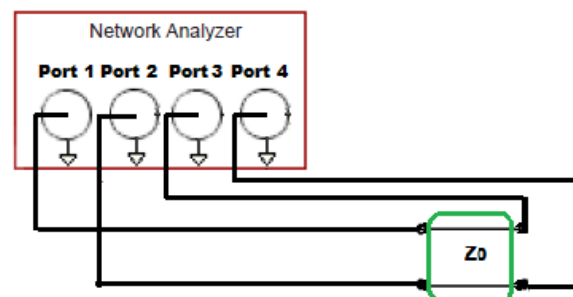
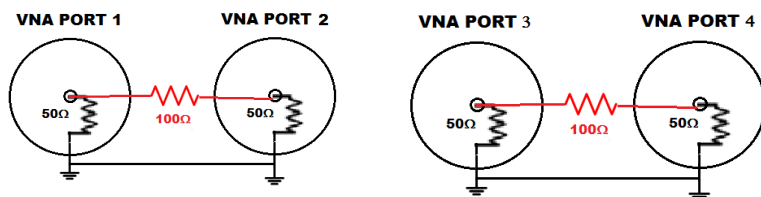
VNA Configurations for Reflection Measurements

Multiport (4 Port) Based



VNA Configurations for Transmission Measurements

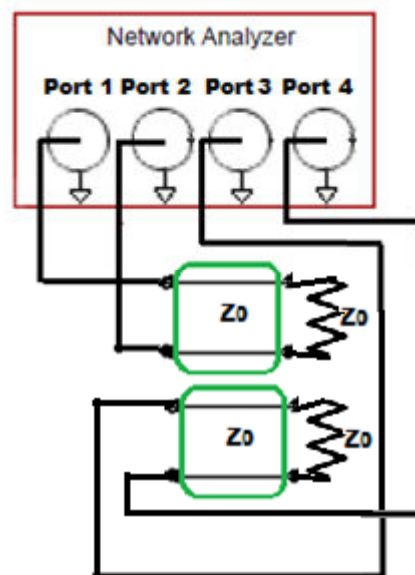
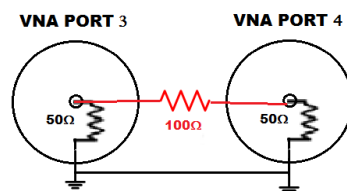
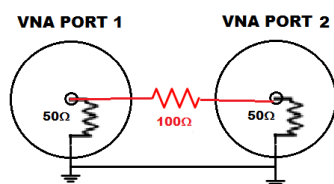
Multiport (4 Port) Based



Use of a Multiport VNA Test Set to Transform 50 Ohm Un-balanced VNA Test Ports to 100 Ohm Balanced Transmission Lines

VNA Configurations for NEXT Measurements

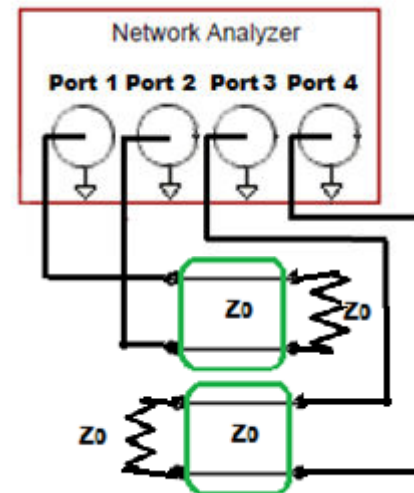
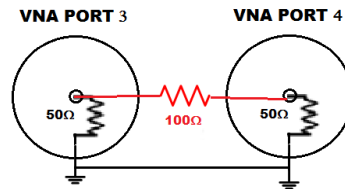
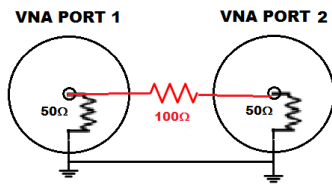
Multiport (4 Port) Based



Use of a Multiport VNA Test Set to Transform 50 Ohm Un-balanced VNA Test Ports to 100 Ohm Balanced Transmission Lines

VNA Configurations for FEXT Measurements

Multiport (4 Port) Based



Advantages and Limitations of Each Approach Balun Based Balanced VNA

Advantages

1. Lowest Cost Balanced VNA Approach
2. True Differential Stimulus
3. Simple Balanced VNA Calibration (SOLT , TRL) With Fewer Connections
4. Traditional Balanced VNA Approach

Limitations

1. Limited Measurement Bandwidth With Transformer Baluns
2. Difficult to Separate Residual Common Mode Component (Longitudinal Balance)
3. Difficult to Measure Mixed Mode and Common Mode S Parameters
4. No Single VNA Vendor Support – “Science Fair Project”
5. VNA Based Reflection and Transmission Time Domain (TDR / TDT) Limited to Band Pass, Impulse Response

Advantages and Limitations of Each Approach 2 Port Based Balanced VNA

Advantages

1. Lowest Cost VNA Configuration
2. Simple Balanced VNA Reflection Calibration (SOLT)
3. Full VNA Reflection Time Domain (TDR) Available

Limitations

1. Reflection Only Balanced VNA Measurements
2. Not True Differential Stimulus (Superposition) With Standard, Lowest Cost 2 Port VNA's

Advantages and Limitations of Each Approach Multiport 4 Port Based Balanced VNA

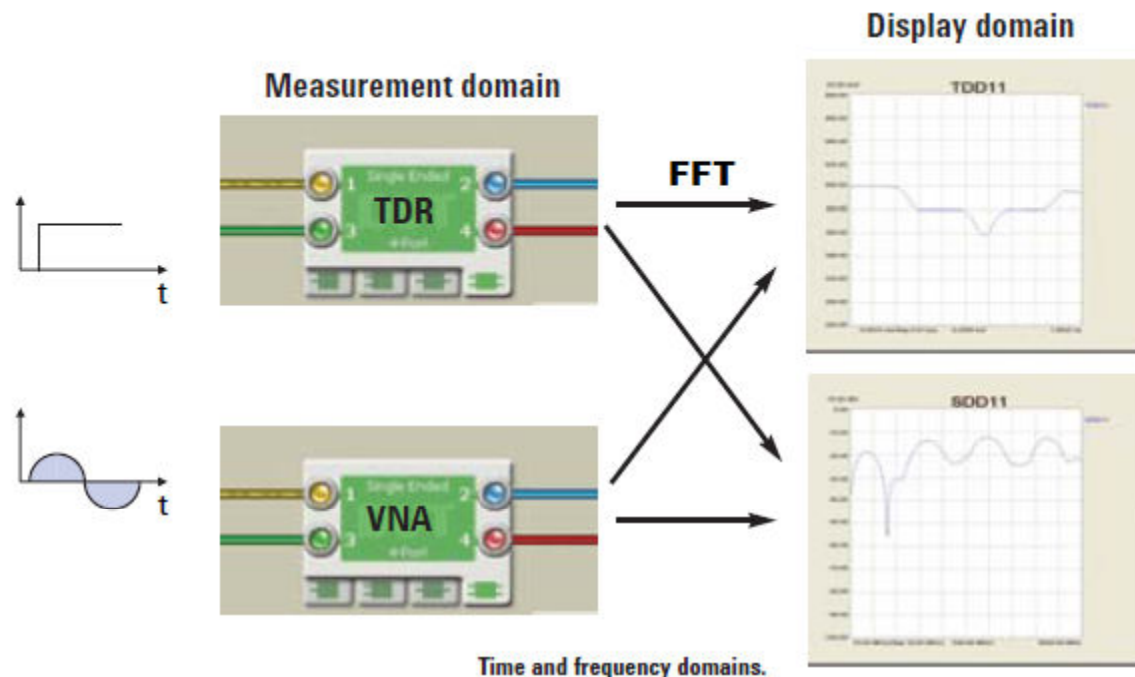
Advantages

1. True Differential Stimulus With Two Coherent Source 4 Port VNA
2. Simple to Measure Differential Mode, Mixed Mode and Common Mode S Parameters
3. Full VNA Reflection and Transmission Time Domain (TDR / TDT) Available

Limitations

1. Not True Differential Stimulus (Superposition) With Standard, Single Source 4 Port VNA's
2. Cumbersome Balanced VNA Calibration (SOLT) With Many Connections
3. Most Expensive Balanced VNA Approach
4. External Test Set Multiport 4 Port Based Balanced VNA Usually Requires External Software for Test Set Control

Frequency Domain Measurement for Balanced Transmission Lines

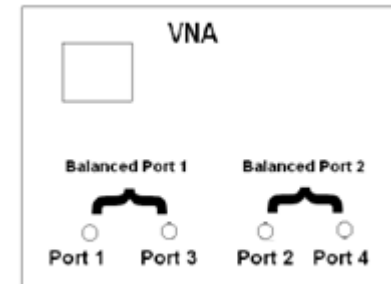
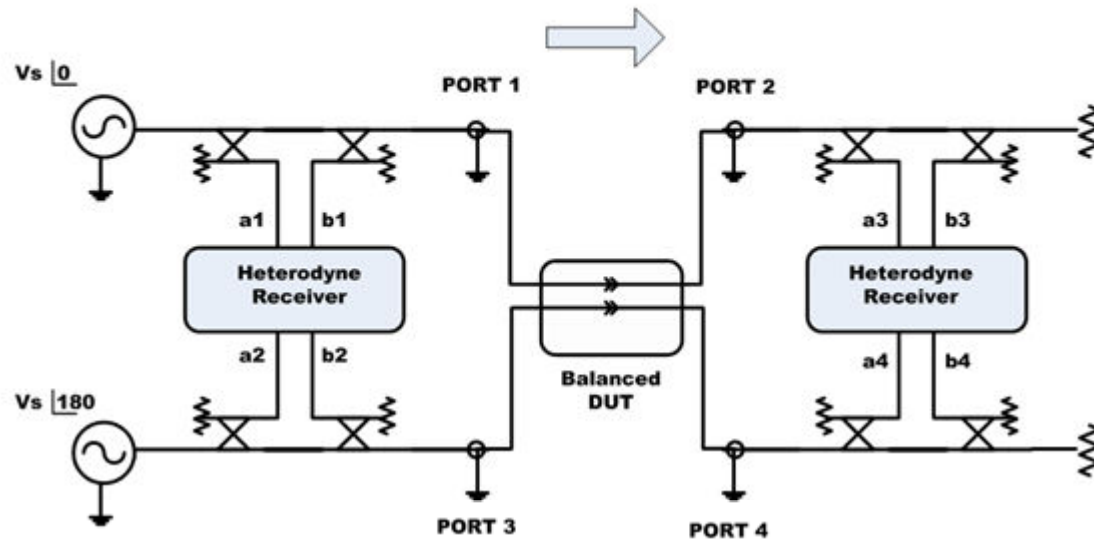


- S Parameters are Complex Numbers (Real and Imaginary) or (Magnitude and Phase)
- Frequency Domain Measurements are Parameter Versus Frequency
- Frequency Domain Measurements Represent a Weighted Average Over Time
- Time Domain Measurements are Parameter Versus Time
- Time Domain Measurements Represent a Weighted Average Over Frequency

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Frequency Domain Measurement with a VNA

Forward Measurements VNA Configuration

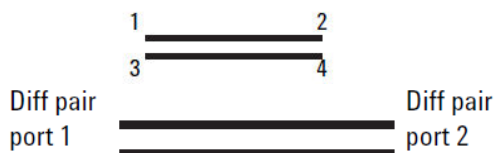


PARAM.	Meas.	Stim. Port*	Meas. Port*
SD1D1	Return loss	P1 Diff	P1 Diff
SD1D2	Rev Transmission	P2 Diff	P1 Diff
SD1C1	Return loss	P1 Com	P1 Diff
SD1C2	Rev Transmission	P2 com	P1 Diff
SD2D1	Fwd Transmission	P1 Diff	P2 Diff
SD2D2	Return loss	P2 Diff	P2 Diff
SD2C1	Fwd Transmission	P1 Com	P2 Diff
SD2C2	Return loss	P2 Com	P2 Diff
SC1D1	Return loss	P1 Diff	P1 Com
SC1D2	Rev Transmission	P2 Diff	P1 Com
SC1C1	Return loss	P1 Com	P1 Com
SC1C2	Rev Transmission	P2 Com	P1 Com
SC2D1	Fwd Transmission	P1 Diff	P2 Com
SC2D2	Return loss	P2 Diff	P2 Com
SC2C1	Fwd Transmission	P1 Com	P2 Com
SC2C2	Return loss	P2 Com	P2 Com

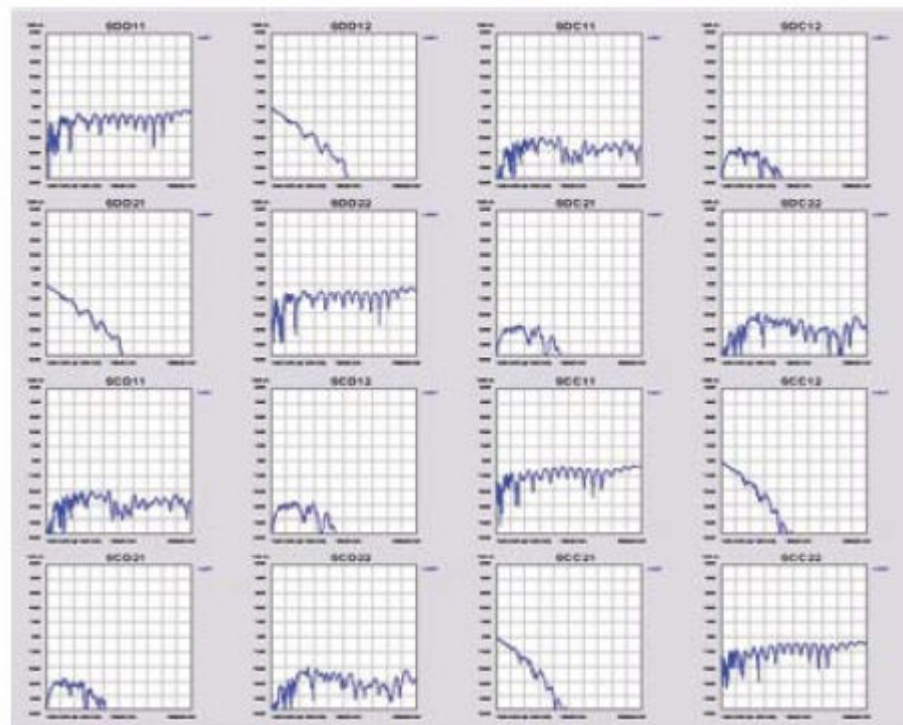
- Stimulus Sources are Sine Wave Oscillators that are Stepped in Frequency
- All Receiver Channels are Narrowband Tuned Receivers Synchronized to the Stimulus Sources
- S Parameters are Complex Numbers (Real and Imaginary) or (Magnitude and Phase)
- Frequency Domain Measurements are Parameter Versus Frequency
- Frequency Domain Measurements Represent a Weighted Average Over Time

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Frequency Domain Measurement with a VNA



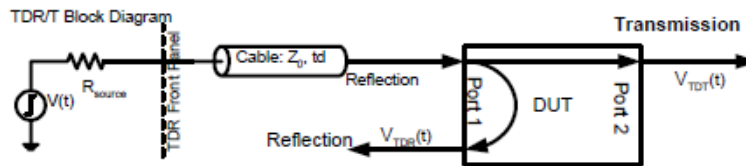
		Stimulus			
		Differential signal		Common signal	
		port 1	port 2	port 1	port 2
Response	Differential signal	port 1	port 2	port 1	port 2
		SDD11	SDD12	SDC11	SDC12
	Common signal	SDD21	SDD22	SDC21	SDC22
		SCD11	SCD12	SCC11	SCC12
	Common signal	SCD21	SCD22	SCC21	SCC22



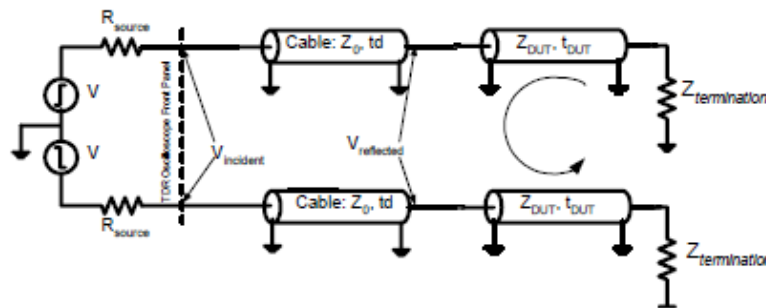
Mixed Mode S Parameters

- SDD are Differential (Balanced) In and Out
- SCC are Common (Single-Ended) In and Out
- SDC are Differential (Balanced) In and Common (Single-Ended) Out
- SCD are Common (Single-Ended) In and Differential (Balanced) Out

Frequency Domain Measurement with a TDR / TDT (FFT)



Time Domain Reflection and Transmission (TDR and TDT) block diagram. A similar diagram can be drawn for reverse measurements (from port 2 to port 1).



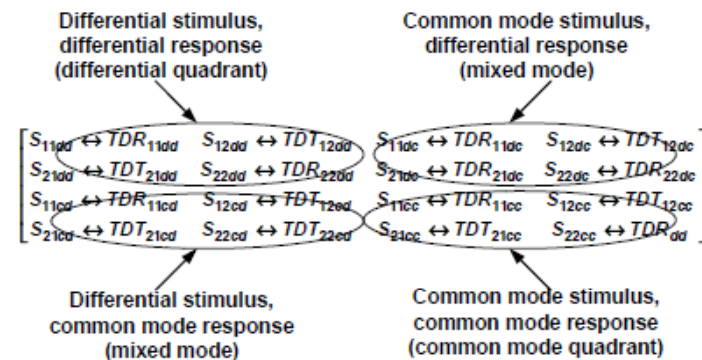
A differential, common mode or mixed mode measurement would require a 4-port measurement setup

$$TDR: \rho = \frac{V_{reflected}}{V_{incident}} = \frac{Z_{load} - Z_0}{Z_{load} + Z_0}$$

$$Z_{DUT} = Z_0 \cdot \frac{1 + \rho}{1 - \rho}$$

$$VNA: S_{11} = \frac{V_{reflected1}}{V_{incident1}} = \frac{Z_{input(DUT)} - Z_0}{Z_{input(DUT)} + Z_0}$$

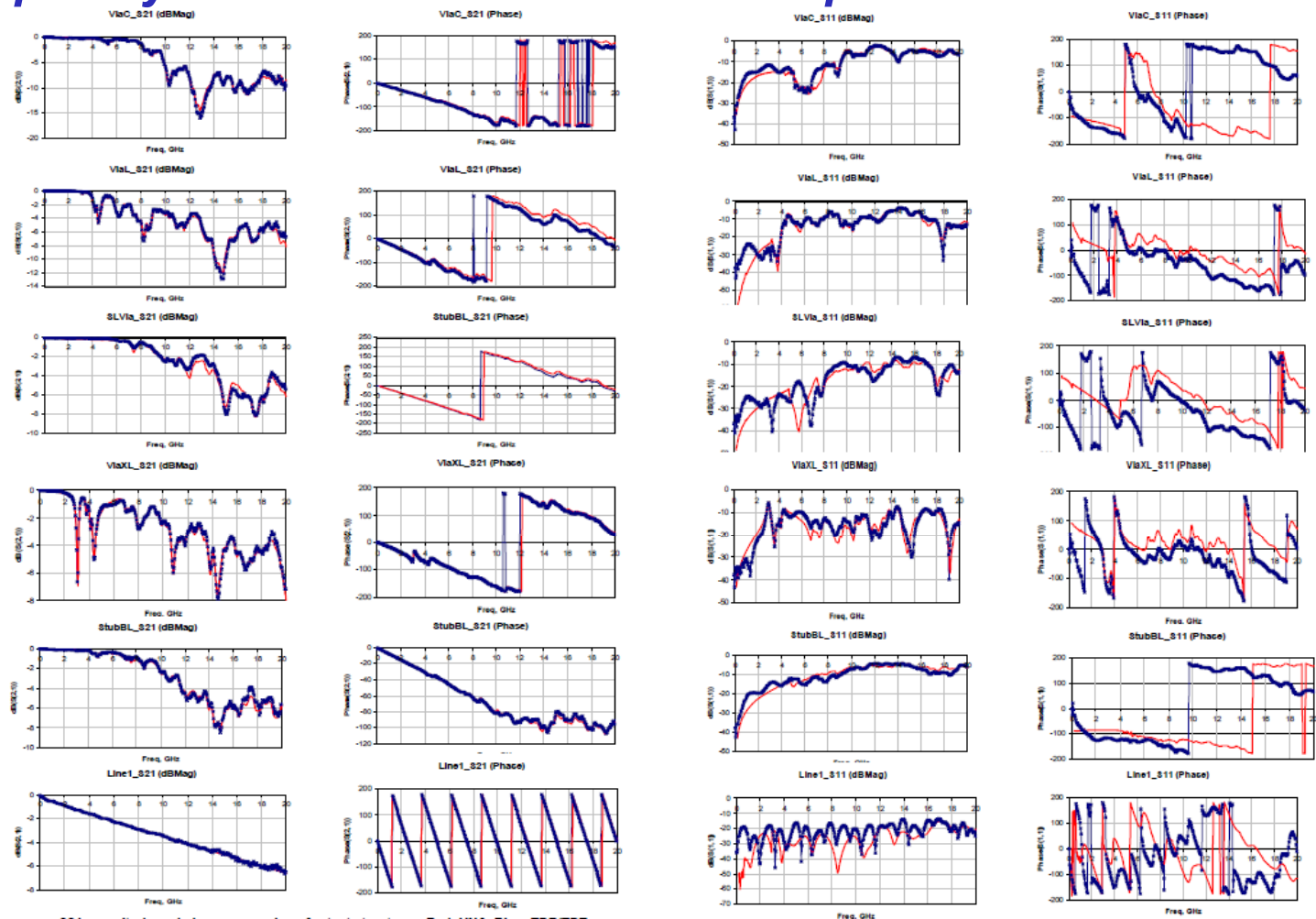
$$Z_{input(DUT)} = Z_0 \cdot \frac{1 + S_{11}}{1 - S_{11}}$$



Correspondence between time and frequency domain waveforms

- Measurements Made With Fast Rise Time Voltage Step Generator
- Both Reflection (TDR) and Transmission (TDT) Made in Time Domain
- Step Response Differentiated to Calculate Impulse Response
- Chirp Z Fast Fourier Transform (FFT) Applied to Impulse Response to Calculate Frequency Domain S Parameters

Frequency Domain Measurement Comparison VNA vs TDR / TDT



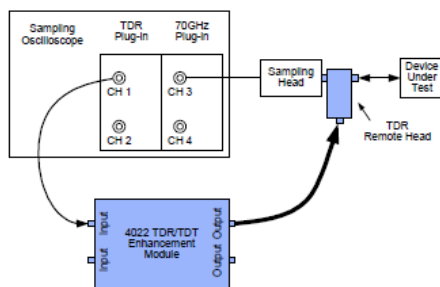
S21 magnitude and phase comparison for test structures. Red: VNA, Blue: TDR/TDT

S11 magnitude and phase comparison for test structures. Red: VNA, Blue: TDR/TDT

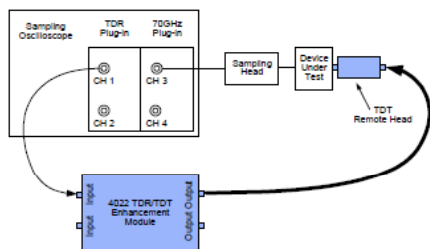
Correlation between VNA and TDR/TDT Extracted S-Parameters
up to 20 GHz

Cherry Wakayama, Jeff Loyer
Intel Corporation, The University of Washington, wakayama@ee.washington.edu
Intel Corporation, jeff.loyer@intel.com

Frequency Domain Measurement Comparison VNA vs TDR / TDT



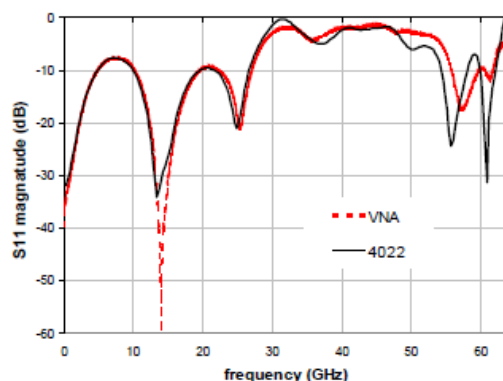
PSPL Model 4022 TDR Test Configuration Used to Measure S11 Data



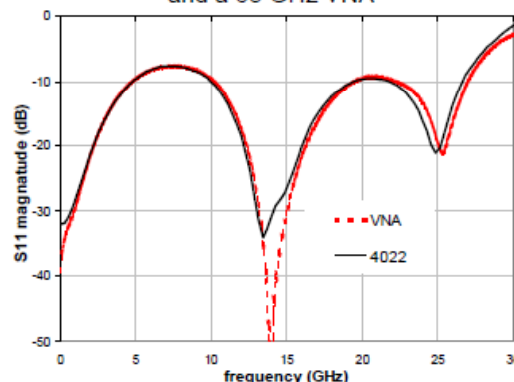
Model 4022 TDT Test Configuration Used to Measure S21 Data

System	S-parameter bandwidth; dyn. range	TDR/T risetime
65 GHz VNA	65 GHz; >60 dB	14 ps
Model 4022 10 ps TDR	30 GHz; >25 dB	10 ps

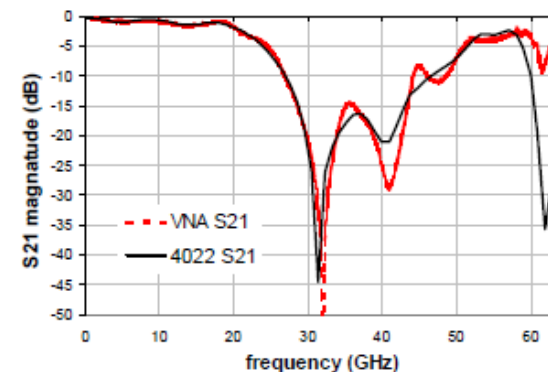
Comparison of TDR/T and VNA Measurement System Capabilities



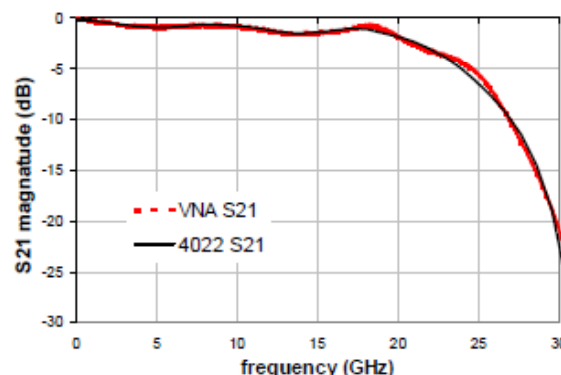
0-64 GHz S11 Plot Measured with the PSPL Model 4022 TDR System and a 65 GHz VNA



0-30 GHz S11 Plot Measured with the PSPL Model 4022 TDR System and a 65 GHz VNA



0-64 GHz S21 Plot PSPL TDR System and 65 GHz VNA Measurements



0-30 GHz S21 Plot PSPL TDR System and 65 GHz VNA Measurements

S-parameter Measurements with the PSPL Model 4022 High-Speed TDR and TDT System

Kipp Schoen, Picosecond Pulse Labs (PSPL), Boulder, Colorado, USA

Advantages and Limitations of Each Approach ***Frequency Domain Measurement with a VNA***

Advantages

1. Higher Source Power and Tuned Receiver → High Dynamic Range (> 100 dB)
2. Applies to Both Active and Passive Linear Devices
3. Direct Frequency Domain Measurements → No Post Processing Required
4. Traditional Measurement Approach For Frequency Domain Performance
5. Better Signal to Noise → Measurements More Repeatable, Less Deviation

Limitations

1. All Measurements are Causal → No Insight Into DUT Topology
2. Higher Cost Than Comparable TDR/TDT Oscilloscope
3. Slower Measurement Speed Than Comparable TDR/TDT Oscilloscope
4. More Complex, Less Bandwidth Signal Separation Test Set

Advantages and Limitations of Each Approach Frequency Domain Measurement with a TDR/TDT

Advantages

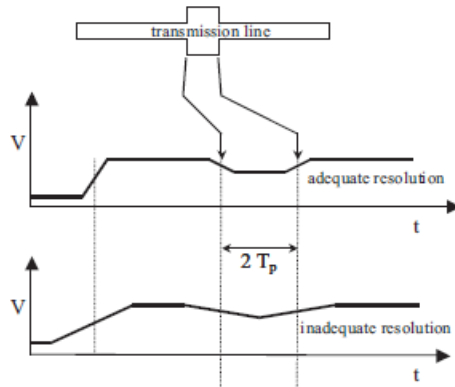
1. Wide Measurement Bandwidth → DC to Max Frequency
2. Lower Cost Stimulus (Step Generator(s))
3. Lower Cost Per Measurement Channel
4. Broadband Receiver → Measurements Made Faster, Single Shot Capture

Limitations

1. Applies to Only Passive Devices
2. Limited Dynamic Range → 40 dB to 50 dB
3. Limited Power in Harmonics of Voltage Step Generator → Less Signal to Noise
4. Direct Time Domain Measurements → FFT Post Processing Required
5. Additional Software Required For Frequency Domain Performance

Time Domain Characteristics and Parameters for Balanced Transmission Lines

TDR Spatial Resolution Requirements



Rise time, ps	Resolution in air, mm	Resolution in FR4, buried run ($v_p=0.446 \cdot C_{light}$), mm
10	1.50	0.67
15	2.25	1.00
20	3.00	1.34
28	4.20	1.87
40	6.00	2.68
150	22.50	10.04

Rise Time Requirements for Serial Standards

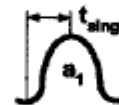
Standards	Datarates, Gb/s*	$t_{\text{standard rise}}$ as a ratio of bit width
1 st generation standards	1.125 - 3.125	15%
2 nd generation standards	4.25 - 6.5	20%
3 rd generation standards	8 - 12	25%

Rise time as percentage of the bit width for three generation of standards.

Faster pulse risetime = better resolution
TDR resolution rules of thumb.



To resolve a_1 and a_2 as
 separate discontinuities:
 $t_{\text{separate}} > t_{\text{TDR_risetime}} / 2$



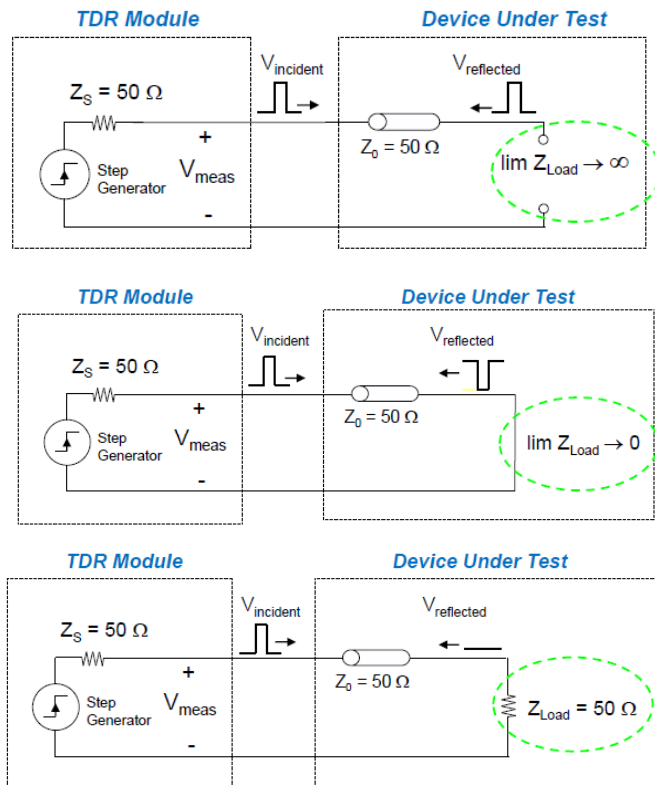
a_1 is not observed if
 $t_{\text{single}} \ll t_{\text{TDR_risetime}}$
 $(t_{\text{single}} < t_{\text{TDR_risetime}} / 10)$

With standard TDR:

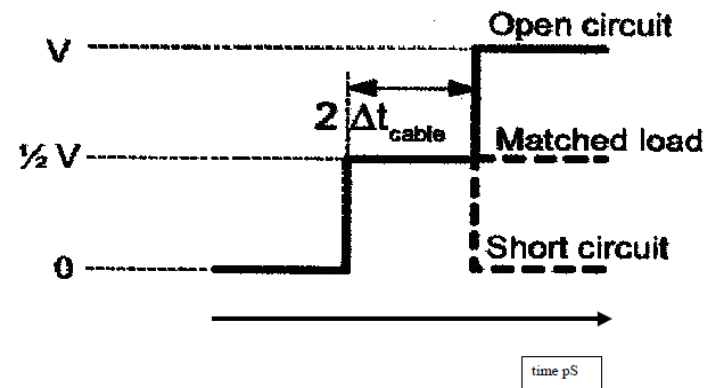
$(t_{\text{separation}} > \frac{1}{2} \text{ TDR risetime} = 17.5\text{ps})$

$(1/10 \text{ TDR risetime} = 3.5\text{ps})$

Time Domain Characteristics and Parameters for Balanced Transmission Lines Impedance



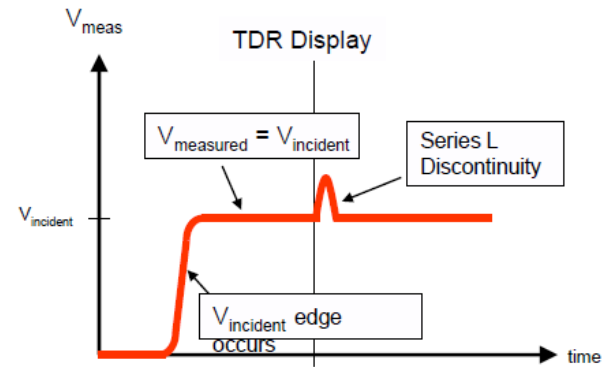
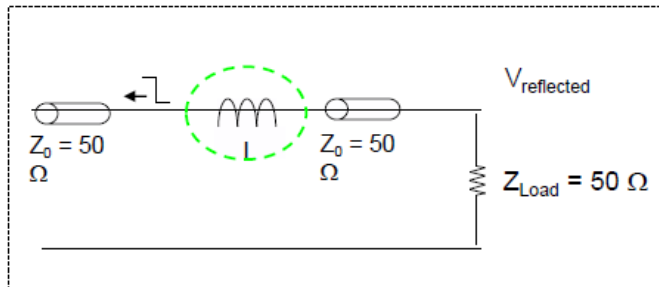
Measurement of impedance from Reflection Coefficient.



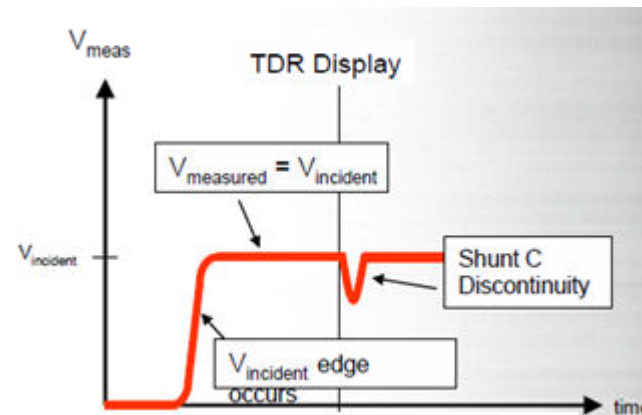
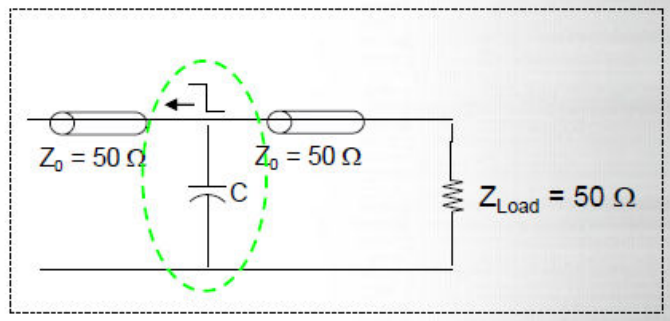
Time Domain Characteristics and Parameters for Balanced Transmission Lines

Fault Type and Location

Device Under Test

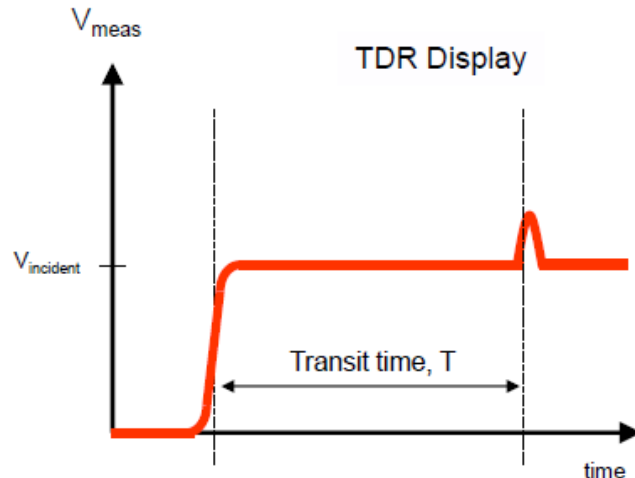
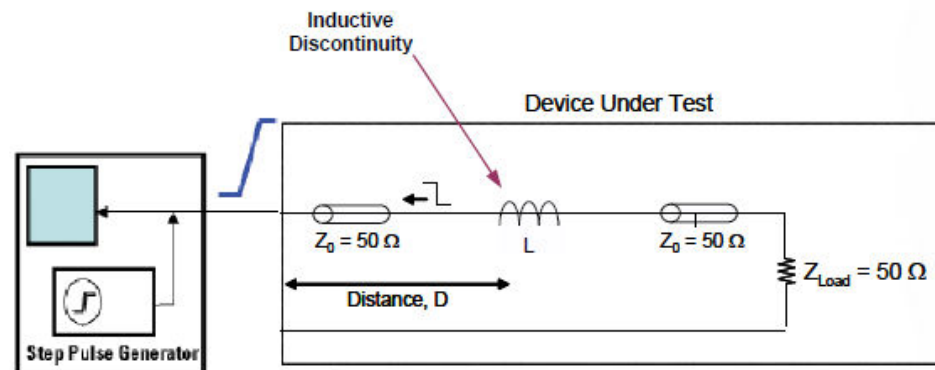


Device Under Test



Time Domain Characteristics and Parameters for Balanced Transmission Lines

Determining Fault Location



Physical distance to fault location can be determined by:

$$D = 0.5 \cdot (T) \cdot (v_p)$$

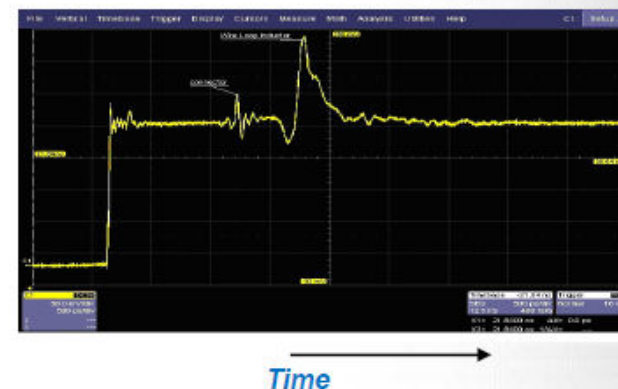
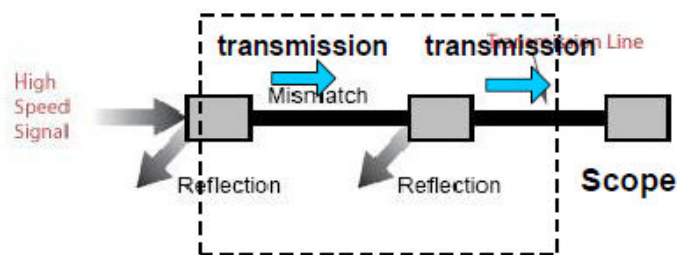
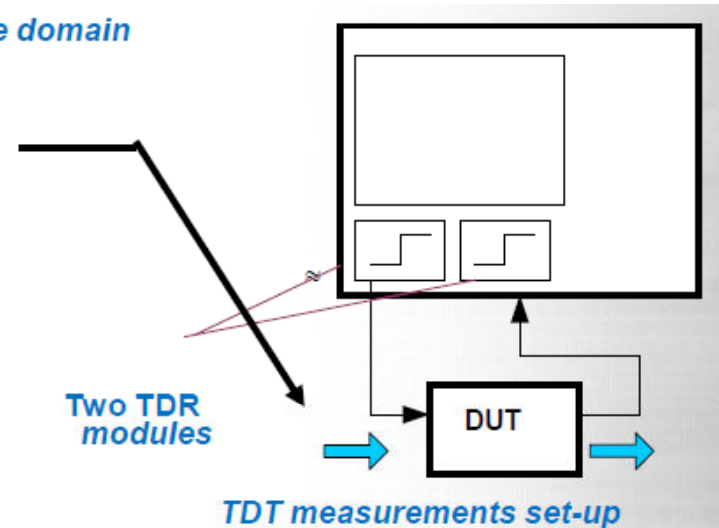
D = physical distance to fault location
 T = transit time from monitoring point to mismatch and back (round trip delay)
 v_p = velocity of propagation (material property)

Time Domain Characteristics and Parameters for Balanced Transmission Lines

TDT - Time Domain Transmission, Insertion Loss, Delay, Skew

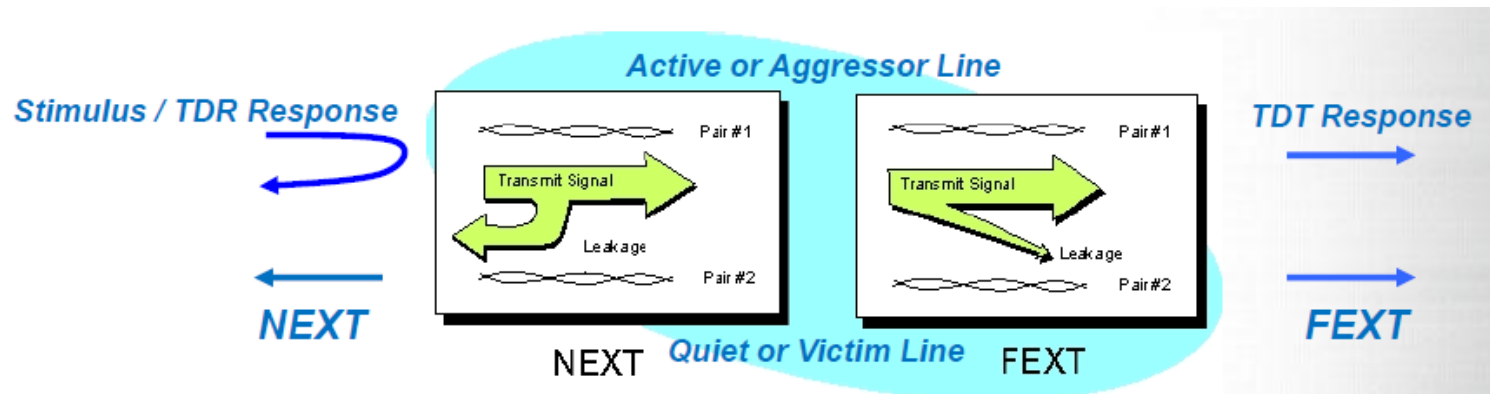
❖ It is the measurements of the transmission in the time domain

- ✓ A pulse generator is used to provide an incident step pulse (stimulus)
- ✓ Voltage Transmission from the Device Under Test (DUT) is measured by the scope
- ✓ $TDT \approx$ Insertion (Transmission) Loss
- ✓ Requires two TDR modules – one to generate the step and other to sample



Time Domain Characteristics and Parameters for Balanced Transmission Lines

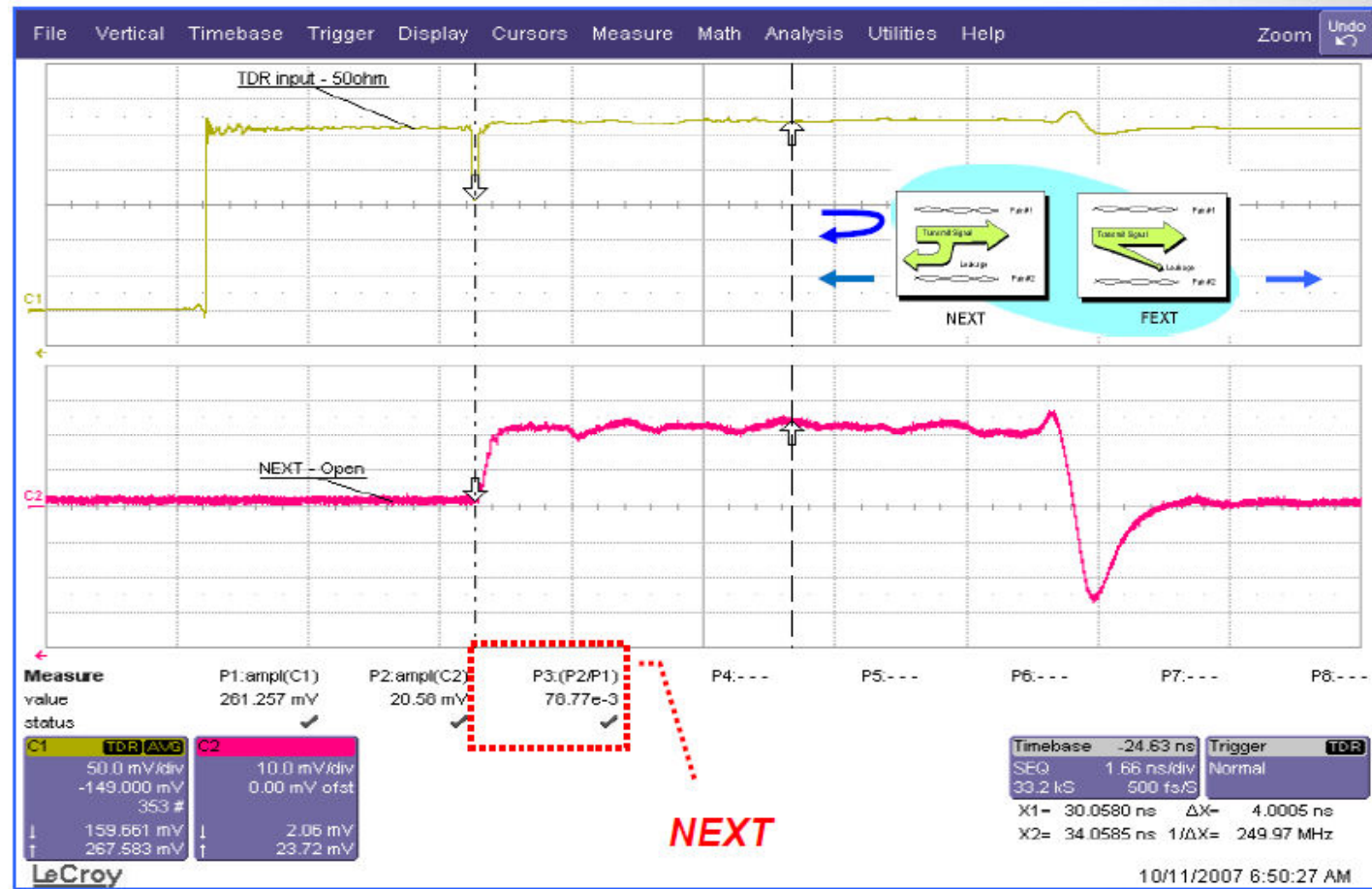
Near End Crosstalk (NEXT), Far End Crosstalk (FEXT)



- ❑ **Near-End Cross Talk (NEXT)** is the ratio between the voltage measured on the near end on the quiet line and the stimulus.
- ❑ **Far-End Cross Talk (FEXT)** is the ratio between the voltage measured on the far end on the quiet line and the stimulus.
- ✓ **NEXT-FEXT Cross-talk** measured the coupling between two adjacent transmission lines

Time Domain Characteristics and Parameters for Balanced Transmission Lines

Near End Crosstalk (NEXT), Quiet Line Open Terminated

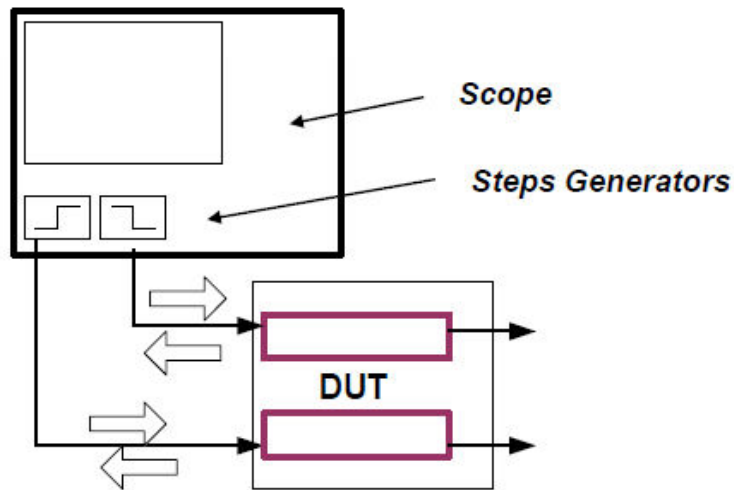


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Time Domain Characteristics and Parameters for Balanced Transmission Lines

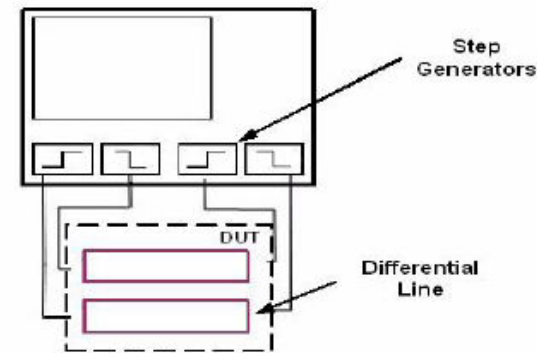
True Differential TDR / TDT

- ✓ High Speed digital systems are mainly differential
- ✓ TDR requires two TDR modules to provide the differential signal (stimulus), step pulses, positive and negative (automatically changes polarity when selecting differential)



Differential TDR measurements set-up

- ✓ De-skew control aligns the two pulses from each of the two TDR modules.
 - ✓ HW deskew (± 50 ps)
- ✓ Requires four TDR modules – two to generate the differential signal and other two to receive the differential signal



Differential TDT measurements set-up

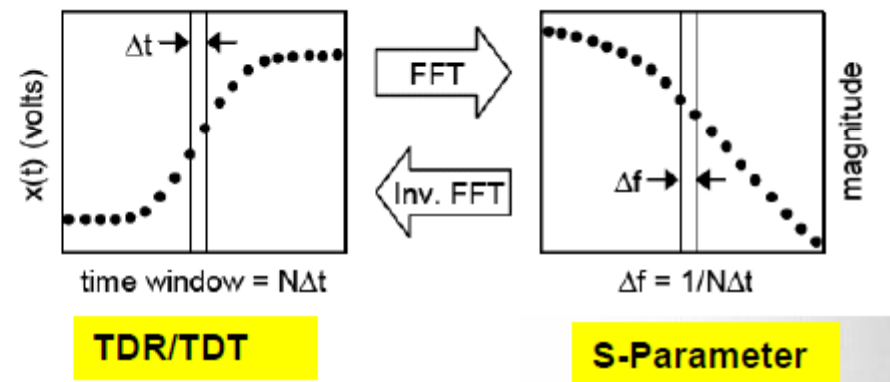
Time Domain Characteristics and Parameters for Balanced Transmission Lines

TDR/TDT and VNA

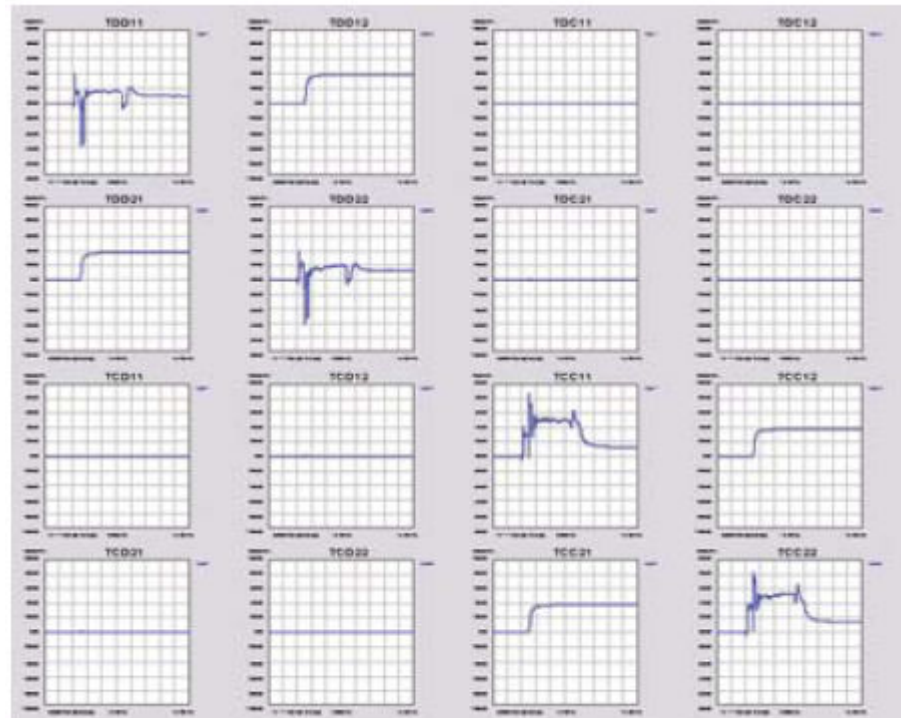
✓ TDR/TDT and S-parameter are describing reflection / transmission respectively in the time domain and in the frequency domain .

✓ TDR/ TDT measurements may be converted into the frequency domain for S-parameter analysis.

✓ S-parameter measurements may be converted into the time domain for TDR/TDT measurements

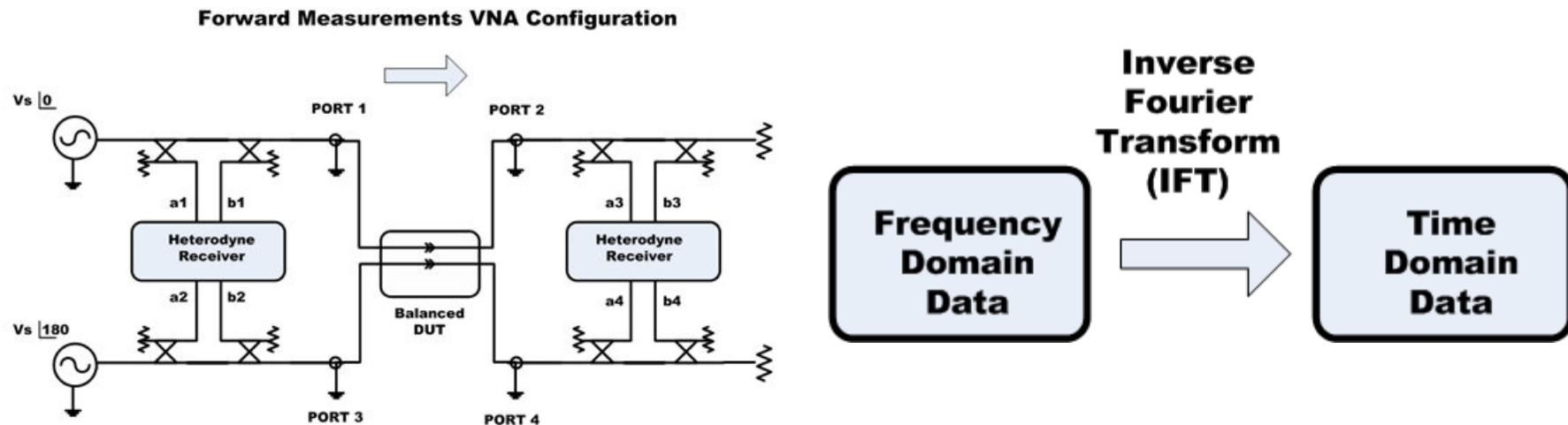


Single Ended, Differential Mode, Common Mode, Mixed Mode, Mode Conversion



- **Time Domain Measurements are Parameter Versus Time / Distance**
- **Time Domain Measurements Represent a Weighted Average Over Frequency**

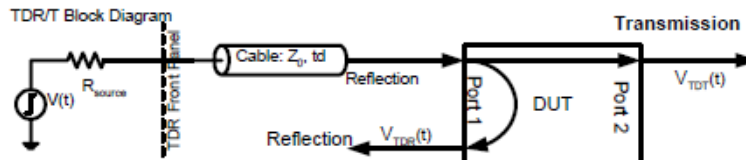
Time Domain Measurement with a VNA (IFT)



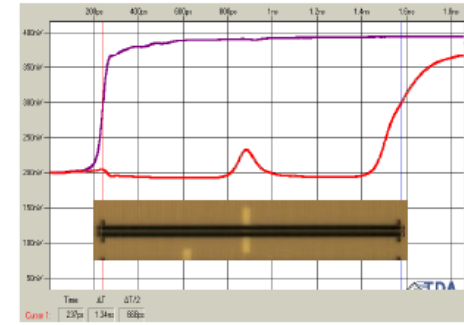
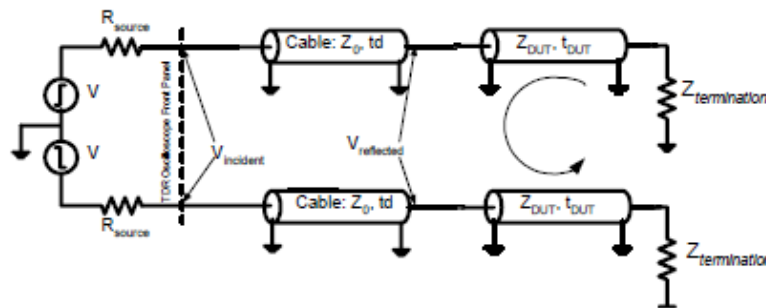
- Measurements Made in Frequency Domain and Converted to Time Domain by Inverse Fourier Transform (IFT)
- Step Response, Impulse Response, and Band Pass Impulse Response
- Step Response and Impulse Response Require Harmonic Related Frequencies
- Time Resolution Related to $(1/\text{Frequency Span})$
- Time Range Related to $(1/\Delta F)$
- Windowing Used to Deal With Frequency Truncation Error (Finite Frequency Range)

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Time Domain Measurement with a TDR / TDT



Time Domain Reflection and Transmission (TDR and TDT) block diagram. A similar diagram can be drawn for reverse measurements (from port 2 to port 1).

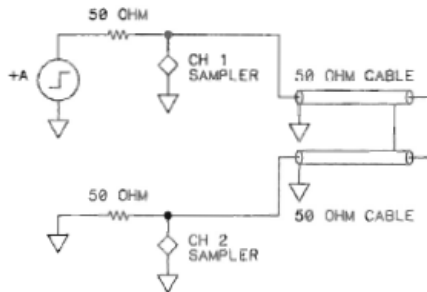


Measured TDR profile of a transmission line with initial via and small gap in the return path.

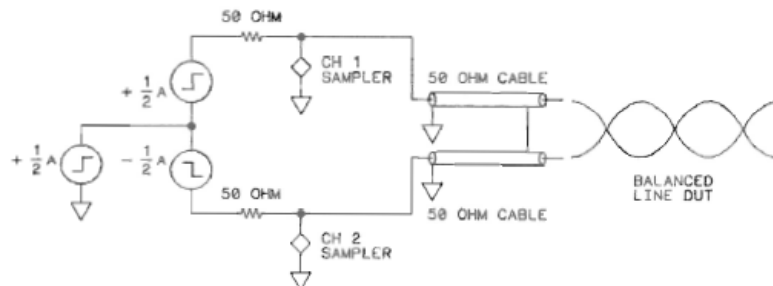
- Measurements Made With Fast Rise Time Voltage Step Generator
- Both Reflection (TDR) and Transmission (TDT) Made in Time Domain
- Step Response Standard For Time Domain Measurements
- TDT Direct Measurement of Insertion Loss, Crosstalk, and Delay
- TDR Direct Measurement of Reflection Coefficient and Impedance with Location

Time Domain Measurement with a TDR / TDT

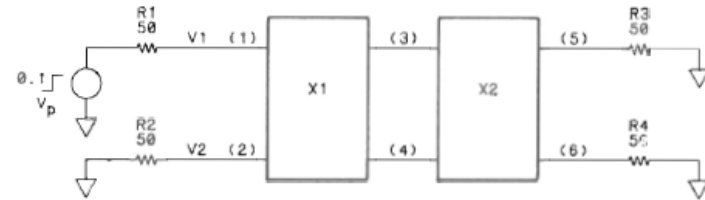
Differential TDR Using a Single Step Generator



Single ended pulse generator built into Ch 1, and Ch 2 input with no step

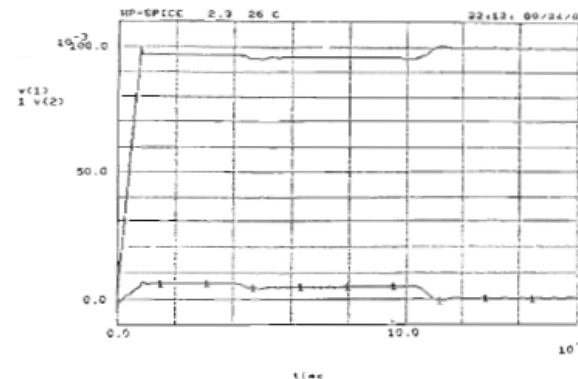


Equivalent circuit of TDR step and scope inputs represented by common mode and differential mode step sources.



Configuration for single ended drive, balanced termination

$Z_0 = 26.5\Omega$
 $\tau_c = 2.12\text{NSEC}$
 $Z_d = 83.27\Omega$
 $\tau_d = 5\text{NSEC}$



V(1) and V(2) responses, single edged drive, balanced termination

- Requires Launch Into Balanced Transmission Line
- Complement Step Generated in Transmission Line by Reciprocity
- Stimulus From Transmission Line is True Differential

Advantages and Limitations of Each Approach

Time Domain Measurement with a VNA (IFT)

Advantages

1. Higher Source Power and Tuned Receiver → Faster Effective Rise Time, Better Time Resolution
2. Both Step and Impulse Response Available
3. Impulse Response of High Pass or Band Pass Devices
4. Time Domain Gating Available
5. DC Response of DUT Not Required (DC Blocks Not a Problem)

Limitations

1. Time Domain Requires Software Post Processing (IFT)
2. Higher Cost Than Comparable TDR/TDT Oscilloscope
3. Slower Measurement Speed Than Comparable TDR/TDT Oscilloscope
4. Windowing Required to Deal With Frequency Truncation Error (Finite Frequency Range)

Advantages and Limitations of Each Approach

Time Domain Measurement with a TDR/TDT

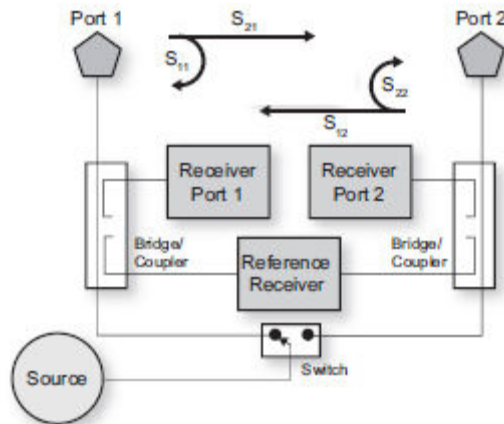
Advantages

1. Direct Time Domain Measurements
2. Direct Time Base Based Delay Measurements
3. Frequency Response of Step Generator Not Band Limited, Windowing Not Required
4. Lower Cost Than Comparable VNA
5. Differential Step Generators With Adjustable Skew

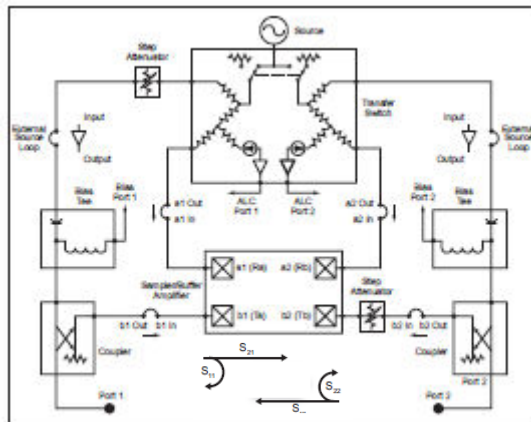
Limitations

1. Step Response Only
2. DC Response of DUT Required (DC Blocks a Problem)
3. Limited Dynamic Range → 40 dB to 50 dB
4. Limited Power in Harmonics of Voltage Step Generator → Slower Rise Time, Less Time Resolution

Techniques Used with MultiPort Vector Network Analyzers (VNA) for Balanced Transmission Line Measurements



Typical 2 Port VNA (3 Receiver)



Typical 2 Port VNA (4 Receiver)

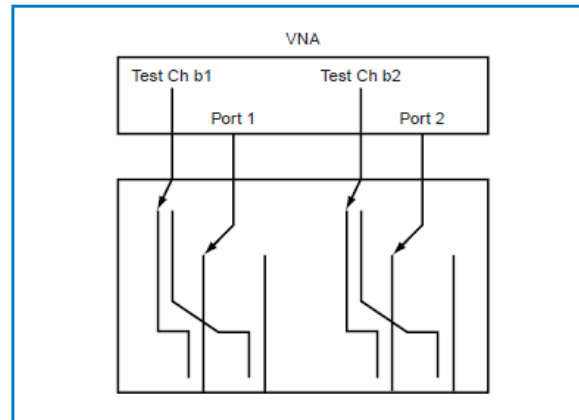


Figure 1. A coupler test set architecture is shown here; the test couplers are on the DUT-side of the multiplexing switches. As N becomes large, this test set becomes quite complex.

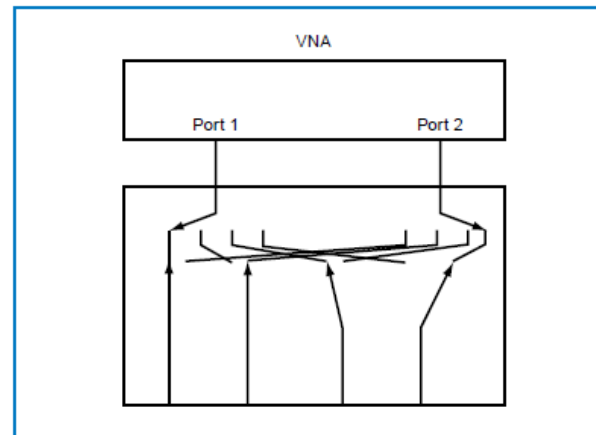
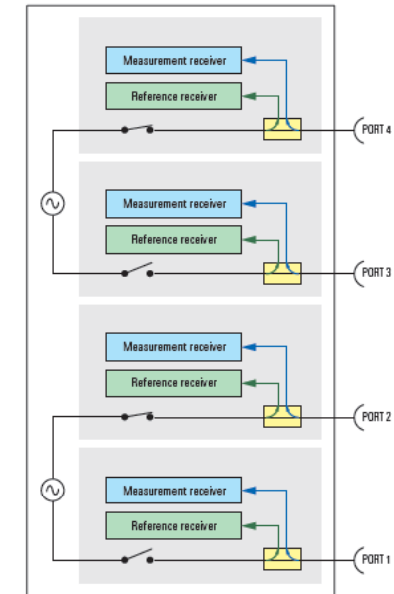
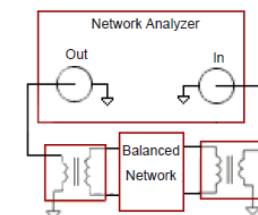


Figure 2. A no-coupler architecture for the 4x2 problem is shown here. In this case, any VNA port can be connected to any test port although this is not needed for most measurements. This test set can be simpler for large N but does have limitations.

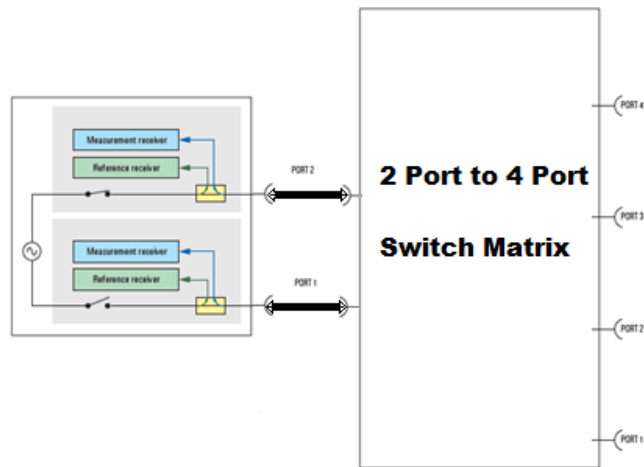


Typical 4 Port VNA (2 Source)



Typical Balun Based VNA

Use of Mathematical Superposition with a Single Source MultiPort VNA for Balanced Transmission Line Measurements

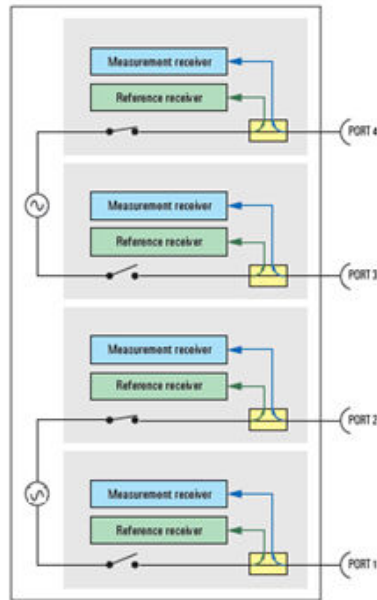


- Passive Balanced / Differential DUT
 - Transmission Lines
 - PCB
 - Lumped Components
 - Passive Filters
 - UTP, STP, Quad Cables
 - Connectors / Interfaces
- Linear Active Balanced / Differential DUT
 - Linear Amplifiers, Differential Amplifiers
 - Linear Active Filters
 - Input / Output Match ADC / DAC

- Standard Single Source VNA Used With Switch Matrix
- Single Ended S Parameters Measured for All Path Combinations
- Differential and Mixed Mode Parameters Calculated by Superposition
- Basic Assumption That DUT Is Linear
- If Interconnect From 4 Port Test Set to DUT Uses Balanced Transmission Lines, True Differential Stimulus Generated By Reciprocity

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Use of Dual Differential Sources MultiPort VNA for Balanced Transmission Line Measurements



- Passive Balanced / Differential DUT
 - Transmission Lines
 - PCB
 - Lumped Components
 - Passive Filters
 - UTP, STP, Quad Cables
 - Connectors / Interfaces
- Linear Active Balanced / Differential DUT
 - Linear Amplifiers, Differential Amplifiers
 - Linear Active Filters
 - Input / Output Match ADC / DAC
- Non Linear Active Balanced / Differential DUT
 - Devices in Compression / Saturation
 - Log Amplifiers

- Dual Source VNA Used
- Sources Are Synchronous and 180 Degree Phase Difference
- Common, Differential and Mixed Mode Parameters Measured Directly
- DUT Can Be Linear or Nonlinear
- True Differential Stimulus
- Balun Based VNA True Differential for Differential Mode Parameters

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Advantages and Limitations of Each Approach

Mathematical Superposition with a Single Source MultiPort VNA

Advantages

1. Lower VNA System Cost
2. If Interconnect From 4 Port Test Set to DUT Uses Balanced Transmission Lines, True Differential Stimulus Generated By Reciprocity
3. Single Ended 2 Port VNA Can Be Upgraded to 4 Port Balanced VNA By Adding Test Set
4. Valid Balanced / Differential Measurements For All Linear DUT's

Limitations

1. Requires Software For Mathematical Superposition Calculation
2. Stimulus Not True Differential * → May Not Be Valid For Non Linear DUT's
3. Switch Matrix Repeatability Can Degrade Systematic Error Correction
Resulting in Higher Residual Errors (Ripple) in Corrected Measurements

Advantages and Limitations of Each Approach Dual Differential Sources MultiPort VNA

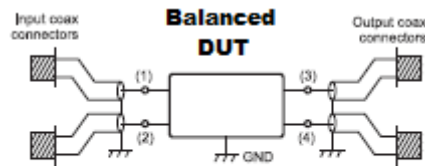
Advantages

1. True Differential Stimulus
2. Valid Balanced / Differential Measurements For All Linear and Non Linear DUT's
3. Better Test Set Repeatability → Lower Residual Errors (Ripple) in Corrected Measurements

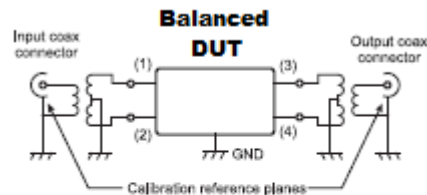
Limitations

1. Higher VNA System Cost
2. 180 Degree Phase Relationship Difficult to Maintain From VNA to DUT → Skew Error → Problematic For Non Linear DUT's

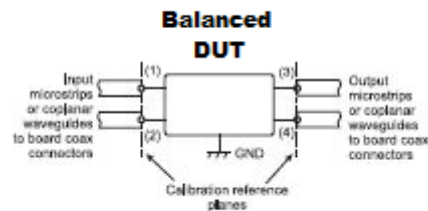
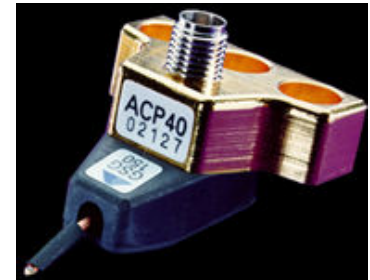
Fixture and Launch Considerations in Connecting to Balanced Structures



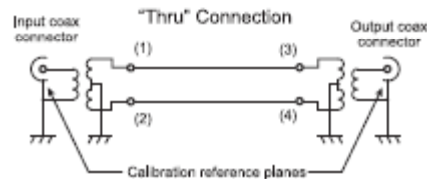
Four-port test jig using short, equal length fixtures.



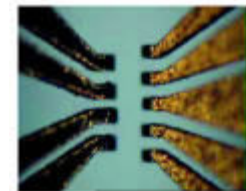
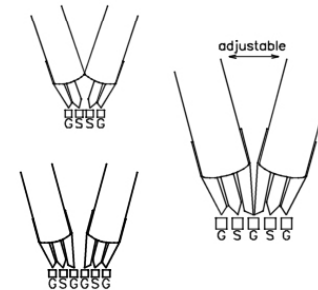
Measurement set-up using input/output baluns



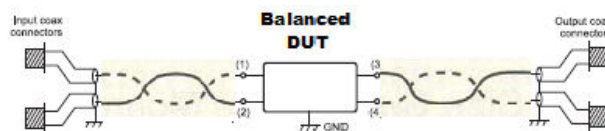
Four-port high-frequency test fixture.



Baluns response calibration



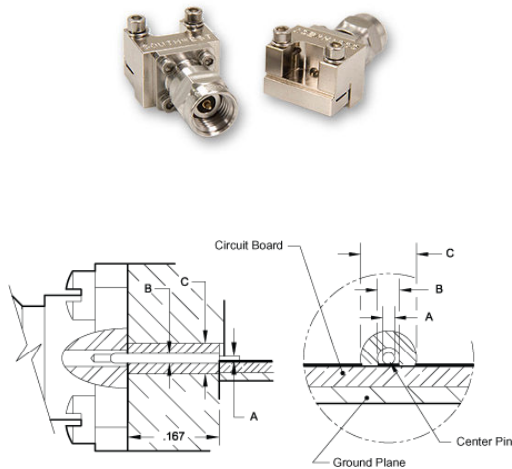
ACP-ESG (left) vs. ACP-ESGSG (right)



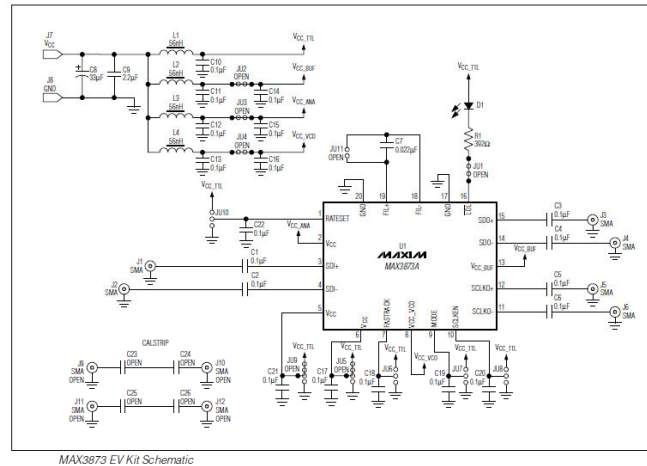
Coax to UTP Transition

Fixture and Launch Considerations in Connecting to Balanced Structures

End Launch Connectors



www.southwestmicrowave.com



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Application Note AN-8

Two-Port Balanced Network Measurements

North Hills™ Signal Processing Corp

Application Note # 160

Agilent Signal Integrity Analysis Series

Part 2: 4-Port TDR/VNA/PLTS

Application Note 5989-5764EN

Time Domain Reflectometry (TDR) and S-parameters

"Advanced Measurementsnot only Signal Integrity" – July 2009

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