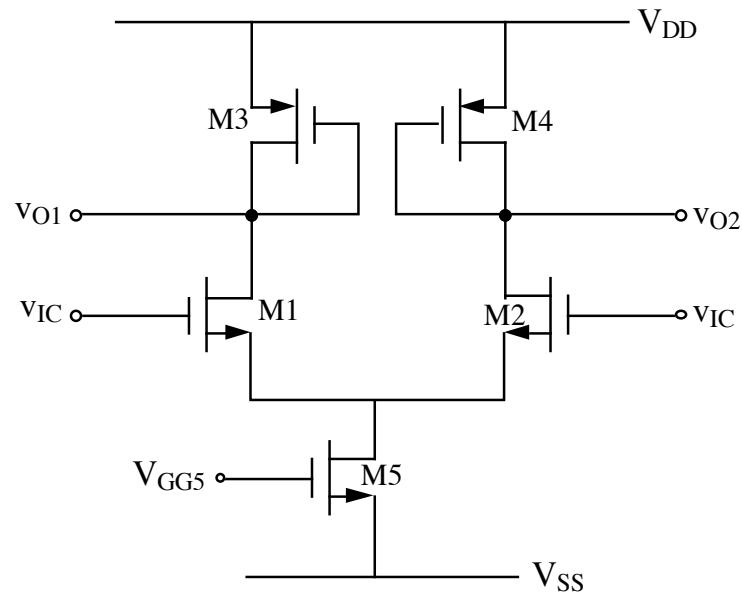
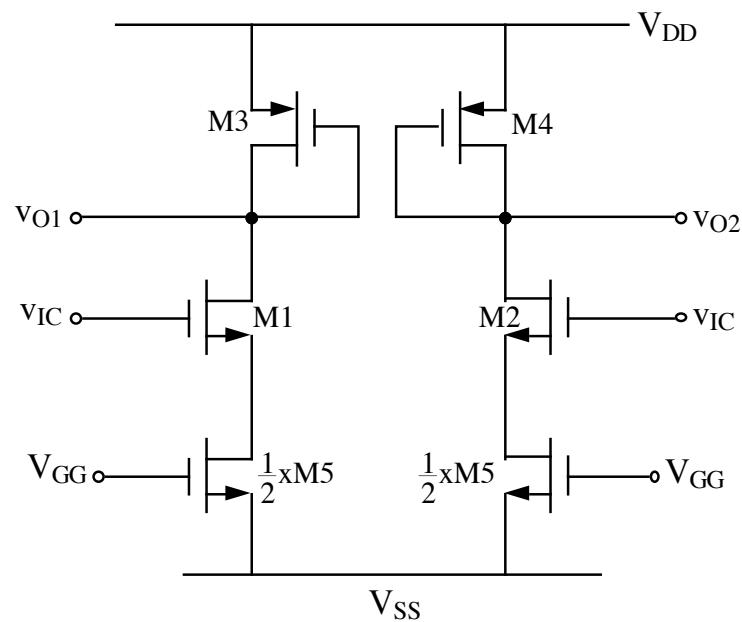


CMOS DIFFERENTIAL AMPLIFIER

Consider the following differential amplifier -

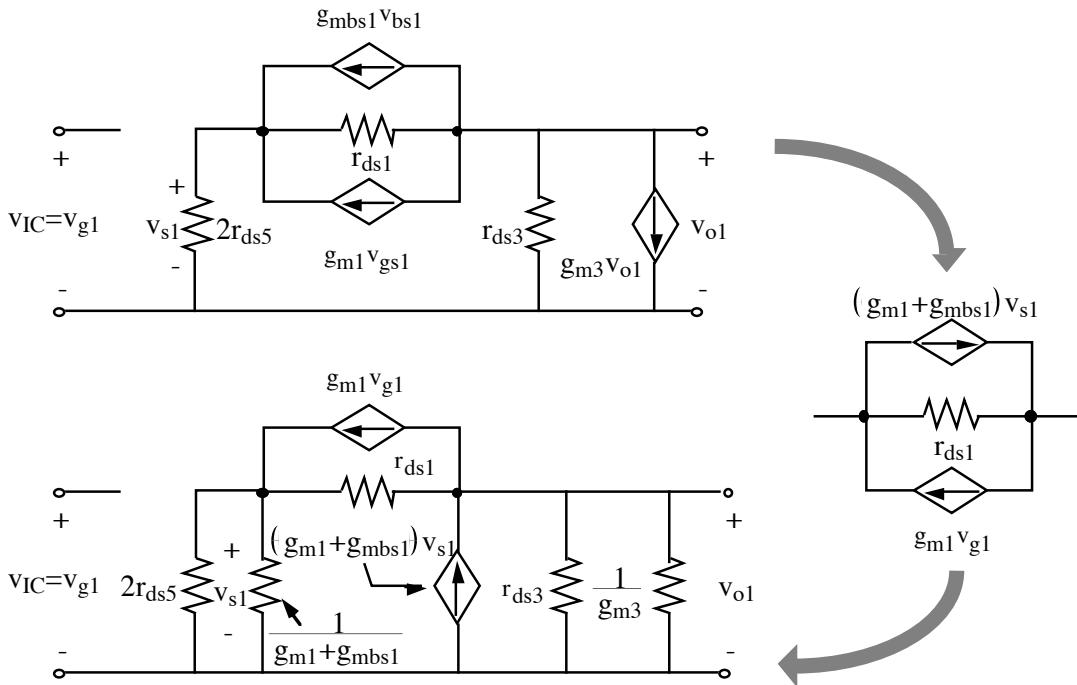


Use of symmetry to simplify gain calculations -



CMOS DIFFERENTIAL AMPLIFIER

Small signal model -



Writing nodal equations -

$$\begin{aligned} [0.5g_{ds} + g_{ds1} + g_{mbs1}] v_{s1} - [g_{ds1}] v_{o1} &= g_m v_{IC} \\ -[g_{ds1} + g_m + g_{mbs1}] v_{o1} + [g_{ds1} + g_{ds3} + g_m] v_{o1} &= -g_m v_{IC} \end{aligned}$$

Solving for $\frac{v_{o1}}{v_{IC}}$ gives,

$$\frac{v_{o1}}{v_{IC}} = \frac{-0.5g_m g_{ds5}}{(g_{ds3} + g_m)[0.5g_{ds} + g_m + g_{mbs1} + g_{ds1}] + 0.5g_{ds1}g_{ds5}}$$

or

$$\frac{v_{o1}}{v_{IC}} \approx \frac{-0.5g_m g_{ds5}}{g_m(g_m + g_{mbs1})} \approx \frac{-g_{ds5}}{2g_m}$$

COMMON MODE REJECTION RATIO (CMRR)

$$\text{CMRR} = \frac{\text{Differential mode gain}}{\text{Common mode gain}} = \frac{A_{vd}}{A_{vc}}$$

For the previous example,

$$|\text{CMRR}| = \frac{\left(\frac{g_{m1}}{g_{m3}}\right)}{\left(\frac{g_{m1}g_{ds5}}{2g_{m3}(g_{m1} + g_{mbs1})}\right)} = \frac{2(g_{m1} + g_{mbs1})}{g_{ds5}} \approx \frac{2g_{m1}}{g_{ds5}}$$

Therefore, current sinks/sources with a larger output resistance(r_{ds5}) will increase the CMRR.

Example

Let all W/L ratios be unity, $I_{SS} = 100\mu\text{A}$, and use the values of Table 3.1-2 to find the CMRR of a CMOS differential amplifier.

$$g_{m1} = \sqrt{2 \times 17 (\mu\text{A}/\text{V}^2) \times 100\mu\text{A}} = 58.3\mu\text{S}$$

$$g_{ds5} = 0.01\text{V}^{-1} \times 100\mu\text{A} = 1\mu\text{S}$$

$$\text{Therefore, } |\text{CMRR}| = 116$$